



HMC690

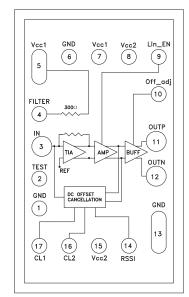
10 Gbps TRANSIMPEDANCE AMPLIFIER

Typical Applications

The HMC690 is ideal for:

- SONET OC-192 and SDH STM-64 Transponders
- 10 Gbps Ethernet
- Broadband instrumentation
- Short, intermediate and long reach optical receiver modules

Functional Diagram



Features

Supports data rates up to 11.3 Gbps 1.25 Kohm differential gain +3.3 V single power supply AC or DC coupled outputs 11 pA/JHz input referred noise density 3 mA p-p overload Average input power monitoring Output offset adjustment Die Size: 0.68 x 1.14 x 0.18mm

General Description

The HMC690 is 10 Gbps transimpedance amplifier designed for SONET OC-192 / SDH STM-64, 10GbE and 10Gbps systems employing optical amplifiers. It supports data rates up to 11.3 Gbps. This amplifier provides a differential output voltage that is proportional to an applied current at its input port. This current is typically provided by a photodiode. Operating from a single +3.3V supply, the HMC690 features low input referred noise, and is designed for driving a CDR or a typical transceiver directly. The RSSI output can be used for monitoring average input power. This device also features a DC offset control, which enables output signal level adjustment for asymmetrical signals. Additional features include an integrated 300Ω filter resistor for photo-diode supply voltage and an extended linear range^[2] option.

Electrical Specifications, $T_A = +25^{\circ}$ C, Vcc1 = Vcc2 = +3.3V^[1]

Parameter	Conditions	Min.	Тур.	Max.	Units
AC Specifications					
Max Data Rate		11.3			Gbps
Small Signal Transimpedance (ZT)	Differential p-p @ 500 MHz	1.1	1.25	1.55	kOhm
Output Amplitude Differential peak-to-peak	For I _{IN} > 1 mA	430	500	560	mV
Small Signal Zt BW	3-dB Upper LImit		7.5		GHz
	Noise Density @ 1 GHz		11.2	13.5	pA/ √ Hz
Input Referred Noise Density	Noise Density @ 5 GHz		13.4	14.7	pA/ √ Hz
Input Referred RMS Noise	@ 8 GHz Bandwidth		1.0		uA rms
Added p-p Deterministic Jitter				10	ps
Random Jitter	l _{IN} > 1 mA		300	500	fs rms
Rise Time	20 - 80%	17	21	25	ps

[1] Lin_en Open

[2] Please see Pin Description table for further explanation of the extended linear range option available on Pin 9.

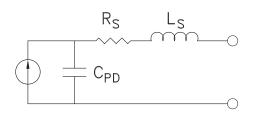


Electrical Specifications (Conditions)

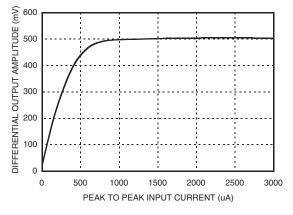
Parameter	Conditions	Min.	Тур.	Max.	Units
Fall Time	80% - 20%	19	22	28	ps
Output Return Loss	F<10 GHz	10			dB
Zt Group Delay Variation			±20		ps
Linear Input Range			300		uA p-p
Input Overdrive			3		mA p-p
Optical Sensitivity ^[1]	@ 10 Gbps (P = 0.9A/W, re = 9 dB, BER = 1e-12)		-20		dBm
DC Specifications					
Power Supply		3.0	3.3	3.6	V
Supply Current		77	92	106	mA
OFF ADJ Sensitivity			780		mV/V
RSSI Sensitivity			-1		mA/V
Open Circuit Input DC Level		0.95		1.1	V

[1] Optical receiver sensitivity depends on packaging, photodiode type, BER value and input signal eye quality.

Photodiode Specification Assumptions



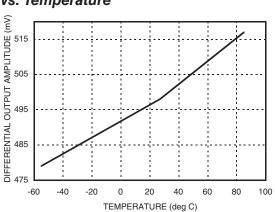
Differential Output Voltage vs. Input Current



Photodiode Capacitance: (Cpd) = 220 fF Photodiode and bond wire parasitic inductance: (Ls) = 800 pH

Photodiode Resistance: (Rs) = 20 Ohm

Photodiode Responsivity: (p) = 0.8 A/W

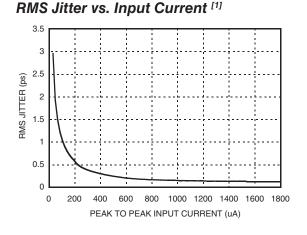


Differential Output Voltage vs. Temperature

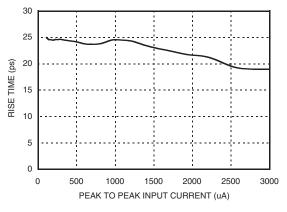
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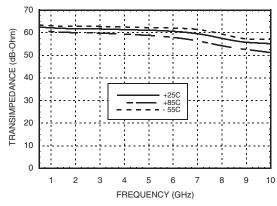




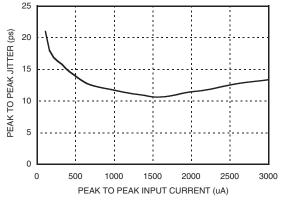
Rise Time vs. Peak to Peak Input Current



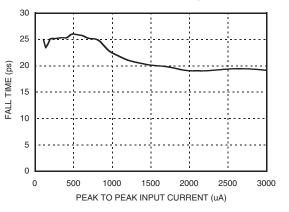
Transimpedance vs. Frequency Over Temperature ^[3]



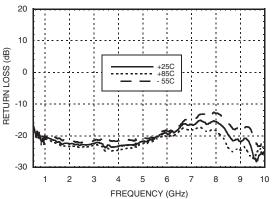
Peak to Peak Jitter vs. Peak to Peak Input Current [2]



Fall Time vs. Peak to Peak Input Current



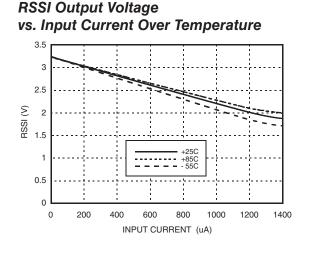




[1] Measured with 10 Gbps 10101 pattern with an estimated bondwire parasitic inductance of 1 nH, source jitter not de-embedded.
[2] Measured with PRBS 2¹⁵⁻¹ pattern at 10 Gbps with an estimated bondwire parasitic inductance of 1 nH. Source jitter not de-embedded.
[3] Single-Ended OUTN

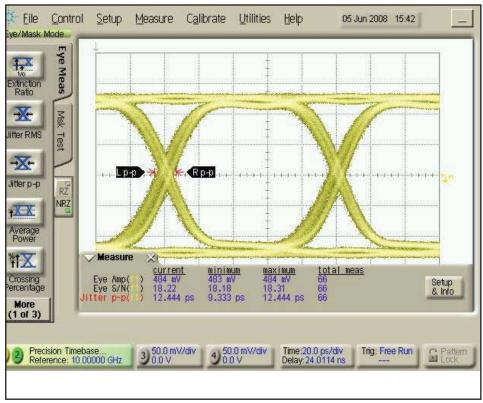
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Output Offset Voltage vs. DC Offset Control Voltage Over Temperature 0.6 0.4 S OUTPUT OFFSET 0.2 0 .250 +25C +85C - 55C -0.2 -0.4 -0.6 1.6 1.8 2.2 2.4 2.6 2.8 3 3.2 2 DC OFFSET CONTROL VOLTAGE (V)

Eye Diagram [4]



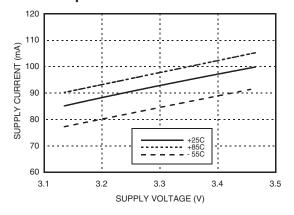
[1] Measured with 10 Gbps 10101 pattern with an estimated bondwire parasitic inductance of 1 nH, source jitter not de-embedded.
[2] Measured with PRBS 2¹⁵⁻¹ pattern at 10 Gbps with an estimated bondwire parasitic inductance of 1 nH. Source jitter not de-embedded.
[3] Single-Ended OUTN

^[4] Output Eye measured on Eval board with 1 mA p-p input current. (10 Gbps), 25C, 3.3V, 2²3-1 pattern





Supply Current vs. Supply Voltage **Over Temperature**



Optical Sensitivity Calculation

Optical sensitivity is determined from the input-referred rms noise current, I_N. To achieve a bit error rate of 1E⁻¹², the signal-to-noise ratio must be 14:1.

$$S = 10 \log \left(\frac{SNR}{2} \times \frac{I_N}{\boxtimes} \times \frac{r_e + 1}{r_e - 1} \times 1000\right) dBm$$

Where

S

= sensitivity (dBm)

SNR = signal to noise ratio (dB)

= input-referred rms noise current (A) I_N ⊠

= photodetector responsitivity (A/W)

= extinction ratio (dB) r_e

Optical Minimum Output Swing at Sensitivity Limit Calculation

The typical optical sensitivity is -19 dBm. At the input level, the voltage swing at output of the HMC690 is calculated as follows:

$$S_{ave} = 10 \log \left(\frac{OMA}{2} \times \frac{re+1}{re-1} \right)$$

Where

Save = average sensitivity (dBm)

OMA = optical modulation amplitude (Wp-p)

= extinction ratio (dB) r_e



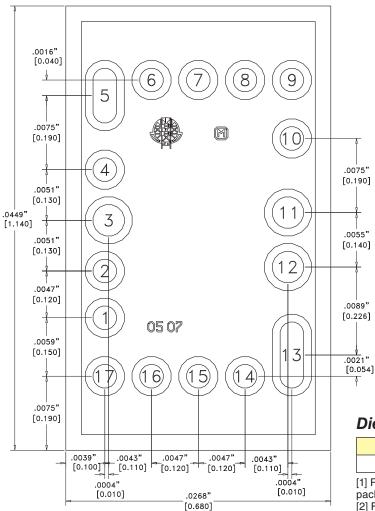
HMC690

10 Gbps TRANSIMPEDANCE AMPLIFIER

Absolute Maximum Ratings

Supply Voltage	4V
Off_adj Voltage	4V
Lin_EN Voltage	4V
Continuous Input Current	8 mA
Junction Temperature	125 °C
Continuous Pdiss (T=85 °C) (derate 9 mW/ °C Above +85 °C	0.36 W
Thermal Resistance (Junction to die bottom)	111 °C/W
Storage Temperature	-65 to 120 °C
Operating Temperature	-40 to +85 °C

Outline



Die Packaging Information ^[1]

Standard	Alternate	
WP-26 (Waffle Pack)	[2]	

[1] Refer to the "Packaging Information" section for die packaging dimensions.

[2] For alternate packaging information contact Hittite Microwave Corporation.





ROHS

10 Gbps TRANSIMPEDANCE AMPLIFIER

Pad Descriptions

Pad Number	Function	Description	Interface Schematic
1, 6, 13	GND	Ground connection for TIA.	
2	TEST	Test Input. This pad is connected internally to IN thru $1k\Omega$. No external connection required.	
3	IN	TIA Input.	TESTO R=1kΩ
4	FILTER	Provides bias voltage for photo diode (PD) thru a 300Ω resistor for Vcc1.	FILTER Vcc 300Ω
5, 7	Vcc1	Power Supply for input stage and PD.	
8, 15	Vcc2	Power supply for output buffers.	
9	Lin_EN	HMC690 has an extended linear range feature. With this feature disabled (pin 9 floating), the HMC690 operates linearly for inputs less than 300μ Ap-p. For input currents greater than 300μ Ap-p, the HMC690 begins to operate within a saturated region. For input currents greater than 1 mA-p-p, the output is fully saturated. Enabling (pin 9 connected to 3.3V) this feature increases linear range of the device up to 350 μ A increasing nominal supply current from 92 mA to 98 mA	Lin_EN 1kΩ
10	Off_adj	DC offset control. Voltage at this pad sets output DC offset. When it is floating DC offset is at 0V.	$\begin{array}{c} & & \\$
11	OUTP	Non-inverted data output with 50 Ω back termination.	ουτρ ουτρ
12	OUTN	Inverted data output with 50 Ω back termination.	
14	RSSI	Received signal strength indicator. This pin provides a voltage proportional to the DC input current. This voltage should be monitored during assembly to optimally align the PD in the optical environment.	

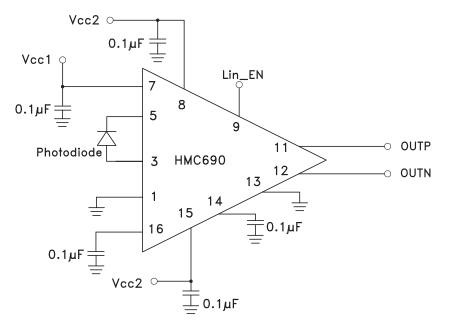
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Pad Descriptions (Continued)

Pad Number	Function	Description	Interface Schematic
16	CL2	Connect a capacitor to ground to increase the on-chip	CL1,20
17	CL1	DC-cancellation loop time constant. 0.1 µF is recommended.	

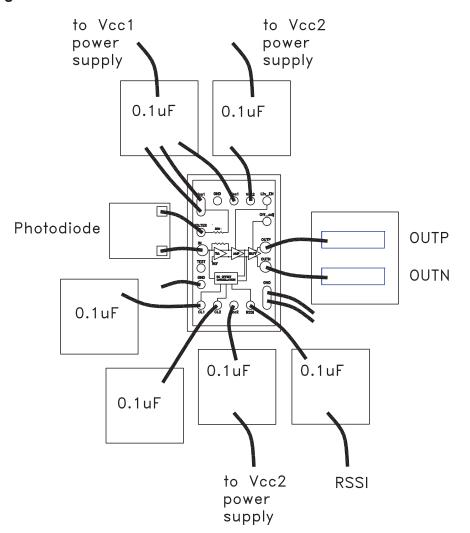
Application Circuit







Assembly Diagram



TRANSIMPEDANCE AMPLIFIERS - CHIP G

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Mounting & Bonding Techniques for Millimeterwave MMICs

The die should be attached directly to the ground plane with epoxy (see HMC general Handling, Mounting, Bonding Note).

50 Ohm Microstrip transmission lines on 0.25mm (10 mil) thick alumina thin film substrates are recommended for bringing RF to and from the chip (Figure 1).

Microstrip substrates should be placed as close to the die as possible in order to minimize bond wire length. Typical die-to-substrate spacing is 0.076mm to 0.152 mm (3 to 6 mils).

Handling Precautions

Follow these precautions to avoid permanent damage.

Storage: All bare die are placed in either Waffle or Gel based ESD protective containers, and then sealed in an ESD protective bag for shipment. Once the sealed ESD protective bag has been opened, all die should be stored in a dry nitrogen environment.

Cleanliness: Handle the chips in a clean environment. DO NOT attempt to clean the chip using liquid cleaning systems.

Static Sensitivity: Follow ESD precautions to protect against ESD strikes.

Transients: Suppress instrument and bias supply transients while bias is applied. Use shielded signal and bias cables to minimize inductive pick-up.

General Handling: Handle the chip along the edges with a vacuum collet or with a sharp pair of bent tweezers.

Mounting

The chip is not back-metallized and should be die mounted with epoxy. The mounting surface should be clean and flat.

Epoxy Die Attach: Apply a minimum amount of epoxy to the mounting surface so that a thin epoxy fillet is observed around the perimeter of the chip once it is placed into position. Cure epoxy per the manufacturer's schedule.

Wire Bonding

RF bonds made with two 1 mil wires are recommended. These bonds should be thermosonically bonded with a force of 40-60 grams. DC bonds of 0.001" (0.025 mm) diameter, thermosonically bonded, are recommended. Ball bonds should be made with a force of 40-50 grams and wedge bonds at 18-22 grams. All bonds should be made with a nominal stage temperature of 150 °C. A minimum amount of ultrasonic energy should be applied to achieve reliable bonds. All bonds should be as short as possible, less than 12 mils (0.31 mm).

