Datasheet

AS1160/AS1161 20MHz - 66MHz, 10-Bit Bus, IEEE 1149.1 (JTAG) Compliant LVDS Serializer/Deserializer

1 General Description

The AS1160 (serializer) is designed to convert 10-bit wide parallel LVCMOS/LVTTL data bus signals into a single high-speed LVDS serial data stream with clock. The AS1161 (deserializer) transforms the high-speed LVDS serial data stream back into a 10-bit wide parallel data bus with recovered parallel clock.

Both devices are compliant with IEEE 1149.1 Standard Test Access Port and Boundary Scan Architecture (including the defined boundary-scan test logic and test access port consisting of Test Data Input, Test Data Out, and Test Mode Select, Test Clock, and Test Reset).

The devices also feature an at-speed BIST mode which allows the interconnects between the serializer and deserializer to be verified at-speed.

The single differential-pair data-path makes PCB design easier, and reduced cable/PCB-trace count and connector size significantly reduce cost. Since one output transmits clock and data bits serially, clock-to-data and datato-data skew are eliminated.

Powerdown mode reduces supply current when both devices are idle.

Both devices are available in a CTBGA 49-bumps pin package.

2 Key Features

- **E** Serial Bus LVDS Data Rate: 660 Mbps @ 66MHz Clock
- 10-bit Parallel Interface
- **E** Synchronization Mode and Lock Indicator
- **Programmable Edge Trigger on Clock**
- **E** High Impedance on Rx Inputs during Poweroff
- **Bus LVDS Serial Output Load: 28** Ω
- **E** IEEE 1149.1 (JTAG) Compliant and At-Speed BIST Test Mode
- **E** Clock Recovery from PLL Lock to Random Data **Patterns**
- Guaranteed Transition each Data Transfer Cycle
- !"Chipset (Tx + Rx) Power Consumption: < 500 mW @ 66MHz
- **E** Single Differential-Pair eliminates Multi-Channel Skew
- **Flow-Through Pinout for Simple PCB Layout**
- Small CTBGA 49-bumps Package

3 Applications

The devices are ideal for cellular phone base stations, add drop muxes, digital cross-connects. DSLAMs, networkswitches and routers or backplane interconnect.

Input Latch DIN0:9 Parallelto-Serial 10

Figure 1. Block Diagrams

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4 Pinout

Pin Assignments and Descriptions

Figure 2. AS1160 Pin Assignments (Top View) $\overline{1}$

A1 DGND	(A3) A2 $D\bar{N}0$ N/C	(A5) AA SYNC ₁ AVDD	AT A6 AVDD N/C	
B1 DIN ₁	(B3) (B2) N/C	(B4) (B5) SYNC2 AVDD AGND	(B6) (B7) AGND AVDD	
$^{\prime}$ C1 DIN ₃	$\left($ C3) $^{\prime}$ C2 $^{\prime}$ DVDD DGND	$^{\prime}$ C5 (C4) DVDD N/C	C7 CG) AGND PWDNN	
(D1) DIN ₅	(D3) (D2) DIN ₄ $D\bar{I}N2$	(D5) (D4) N/C DO-	(D6) (D7) DEN DO+	
E1 DIN ₇	$(\mathsf{E}3)$ (E2) DIN ₆ TMS	(E4) (55) TCLK DVDD	E7 E6 AGND DGND	
F1 TĐI	(F2) (F3) DIN ₈ TCK	(F4) (F5) DIN ₉ DGND	F6 F7 N/C AGND	
G1 TDO	$'$ G2 $'$ (S3) TRSTNTCKR/FN DGND	(G5) (G4) AVDD	$\left($ 67 $\right)$ $($ G6 $)$ N/C N/C	

Table 1. AS1160 Pin Descriptions

Figure 3. AS1161 Pin Assignments (Top View)

(A7) (A5) (AB) (A3) (A4) A2 AA	
DGND REFCLK AGND ROUT1 DGND DVDD N/C	
(B7) (B5) (B6) (B3) (B1) B2) (B4) AGND RCKR/FN ROUT2 DGND ROUT3 DVDD AVDD	
(C5) (C4) (C6) $\left($ C3 $\right)$ C2 $^{\prime}$ C7 $^{\prime}$ (C1) RT- N/C ROUTO DVDD AVDD DVDD ROUT4	
$\left($ D5 $\right)$ (D6) (D7` (D3) (D2) (D4) D1) \widetilde{RI} + PWDNN DVDD ROUT5 DGND REN NTC	
$(\textsf{E4})$ (E3) E 5) E2) E6) E7 E ₁ LOCKN RCLK N/C DGND TCK TRSTN DGND	
(F4) (F3) (F5) F6 (F7) (F1 F2 ROUT ₈ AVDD AGND AGND TDI ROUT ₆ AVDD	
(G7) (G5) (G6) (G3) (G4) (G1 (G2) DGND ROUT9 ROUT7 TMS TĎO AGND AVDD	

Table 2. AS1161 Pin Descriptions

5 Absolute Maximum Ratings

Stresses beyond those listed in [Table 3](#page-4-0) may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in [Electrical Character](#page-5-0)[istics on page 6](#page-5-0) is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

6 Electrical Characteristics

 $AVDD = DVD = 3V$ to 3.6V, TAMB = -40°C to +85°C, R LOAD=28 Ω , CLOAD = 10pF, typical values @ TAMB = +25°C and VDD = 3.3V (unless otherwise specified).

Serializer Electrical Characteristics

Table 4. Electrical Characteristics

Serializer Timing Requirements for TCLK

Table 5. Serializer Timing Requirements for TCLK

Serializer Switching Characteristics

Table 6. Serializer Switching Characteristics¹

1. Guaranteed by simulation and characterization.

2. Because the serializer is in tri-state mode, the deserializer will lose PLL lock and have to resynchronize before data transfer.

Deserializer Electrical Characteristics

 $AVDD = DVD = 3V$ to 3.6V, TAMB = -40°C to +85°C, R LOAD=28 Ω , CLOAD = 15pF, Receiver Input Range: 0V to 2.4V, typical values $@$ TAMB = $+25C$ and V DD = 3.3V (unless otherwise specified).

Table 7. Electrical Characteristics

Table 7. Electrical Characteristics (Continued)

Deserializer Timing Requirements for REFCLK

Table 8. Deserializer Timing Requirements for REFCLK

Deserializer Switching Characteristics

Table 9. Deserializer Switching Characteristics¹

1. Guaranteed by simulation and characterization.

2. For the purpose of specifying deserializer PLL performance, tDSR1 and tDSR2 are specified with the REFCLK running and stable, and with specific conditions for the incoming data stream synchronization patterns (SYNC-PATs). The derserializer should be initialized using either tDSR1 or tDSR2.

tDSR1 is the time required for the deserializer to indicate lock upon power-up or when leaving the power-down mode. SYNCPATs should be sent to the device before initiating either condition.

tDSR2 is the time required to indicate lock for the powered-up and enabled deserializer when the input (RI+ and RI-) conditions change from not receiving data to receiving SYNCPATs.

3. tRNM is a measure of how much phase noise (jitter) the deserializer can tolerate in the incoming data stream before bit errors occur. The deserializer noise margin is guaranteed by design using statistical analysis.

Scan Circuitry Timing Requirements

Table 10. Scan Circuitry Timing Requirements

Note: All limits are guaranteed. The parameters with min and max values are guaranteed with production tests or SQC (Statistical Quality Control) methods.

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 $VDD = 3.6V$, RLOAD = 28 Ω , CLOAD = 10pF, TAMB = +25°C (unless otherwise specified);

8 Typical Operating Characteristics AS1161

 $VDD = 3.6V$, CLOAD = 15pF, TAMB = $+25^{\circ}C$ (unless otherwise specified);

9 Timing Diagrams

Figure 16. Worst-Case Serializer ICC Test Pattern

Figure 17. Worst-Case Deserializer ICC Test Pattern

Figure 20. Serializer Input Clock Transition Time

Figure 21. Serializer Setup and Hold Times

Figure 24. Serializer PLL Lock Time and PWDNN Tri-State Delays

Datasheet - Timing Diagrams

Figure 25. SYNC Timing Delays

Figure 26. Serializer Delay

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Figure 27. Deserializer Delay

Figure 28. Deserializer Data Valid Out Times

Figure 29. Deserializer Tri-State Test Circuit and Timing Diagram

Figure 30. Deserializer PLL Lock Times and PWDNN Tri-State Delays

Figure 31. Deserializer PLL Lock Time from SYNCPAT

Datasheet - Timing Diagrams

Where:

tsw is the setup and hold time (internal data sampling window). tDJIT is the serializer output bit position jitter as a result of jitter on TCLK. tRNM is the receiver noise time margin.

Figure 34. Data Transfer Mode, VOD Diagram (VOD = DO+ - DO-)

10 Detailed Description

The serializer/deserializer chipset transfers 10 parallel LVTTL data bits over a serial Bus LVDS link up to 660Mbps at clock speeds from 20MHz to 66MHz.

For the serializer, an on-board PLL serializes the input data and inserts two control bits (start & stop bit) into the data stream.

The deserializer uses a separate reference clock (REFCLK) and an onboard PLL to extract the clock information from the incoming data stream and then deserialize the data. The deserializer monitors the incoming clock information, determines lock status and asserts the LOCKN output high when loss of lock occurs.

Note: The chipset is also capable of driving data over unshielded twisted pair cable.

The chipset has three active states of operation:

- Initialization
- Data Transfer
- Resynchronization

The chipset also has two passive states:

- Powerdown
- Tri-State

Note: There are also test modes for JTAG access and at-speed BIST (built-in-self-test).

Initialization

Initialization of both devices must occur before data transmission begins. Initialization refers to synchronization of the serializer and deserializer PLLs to local clocks, which may be the same or separate. Afterwards, synchronization of the deserializer to the serializer occurs.

1. When VDD is applied and reaches a stable value between +3.0V and +3.6V, the PWDNN of the serializer has to stay low for at least 50µs to ensure proper operation [\(see Figure 23 on page 14\),](#page-13-2) the respective output enter tri-state. After PWDNN is high the PLL in the serializer begins locking to a local clock.

When V_{DD} is applied to the deserializer, the respective outputs enter tri-state and an on-chip power-on circuitry disables internal circuitry. When VDD reaches VDDOK (2.5V) the PLL in the deserializer begins locking to a local clock. For the serializer, the local clock is the transmited clock (TCLK) provided by the source ASIC or other device. For the deserializer, a local clock must be applied to pin REFCLK that can be provided by any source.

The serializer outputs remain in tri-state while the PLL locks to the TCLK. After locking to TCLK, the serializer is now ready to send data or SYNCPATs (SYNC patterns), depending on the levels of the SYNC1 and SYNC2 inputs or a data stream.

The SYNCPAT sent by the serializer consists of six ones and six zeros switching at the input clock rate. Note that the deserializer LOCKN output will remain high while its PLL locks to the incoming data or to SYNCPATs on the input.

2. The deserializer PLL must synchronize to the serializer to complete initialization. The deserializer will lock to non-repetitive data patterns. However, the transmission of SYNCPATs enables the deserializer to lock to the serializer signal within a specified time [\(see Figure 24 on page 14\)](#page-13-0).

The application determines control of pins SYNC1 and SYNC2. A direct feedback loop from the LOCKN pin is mandatory [\(see Figure 36 on page 23\).](#page-22-0) In all cases the serializer stops sending SYNCPATs after both SYNC inputs return low.

When the deserializer detects edge transitions at the bus LVDS input, it will attempt to lock to the embedded clock information. When the deserializer locks to the bus LVDS clock, the LOCKN output will go low. When LOCKN is low, the deserializer outputs represent incoming bus LVDS data.

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Data Transfer

After initialization, the serializer will accept data from inputs DIN0:DIN9. The serializer uses TCLK to latch incoming data. TCKR/FN selects which edge the serializer uses to strobe incoming data. TCKR/FN high selects the rising edge for clocking data and low selects the falling edge. If SYNC1 or SYNC2 is high for more than 5 TCLK cycles, the data at DIN0:DIN9 is ignored regardless of clock edge.

After determining which clock edge to use a start and stop bit, appended internally, frame the data bits in the register. The start bit is always high and the stop bit is always low. In the serial stream the start and stop bits are used as the embedded clock bits.

The serializer transmits serialized data and clock bits (10 + 2 bits) from the serial data output (DO+ and DO-) at 12 times the TCLK frequency. For example, if TCLK is 66MHz, the serial rate is 66 x 12 = 792 mega bits per second. Since only 10 bits are from input data, the serial payload rate is 10 times the TCLK frequency (if TCLK = 66MHz, the payload data rate is 66 x 10 = 660Mbps). The data source provides TCLK and must be in the range of 20MHz to 66MHz nominal.

The serializer outputs (DO+ and DO-) can drive a point-to-point connection [\(see Figure 37 on page 24\)](#page-23-0) or a multidrop configuration [\(see Figure 35\).](#page-19-0) In a multidrop configuration one serializer is connected through a backplane bus with limited multiple deserializers.

The outputs transmit data when DEN, PWDNN are high and SYNC1, SYNC2 are low.

Note: When DEN is driven low, the serializer output pins will enter tri-state.

When the deserializer synchronizes to the serializer, pin LOCKN is low. The deserializer locks to the internal clock and uses it to recover the serialized data. ROUT0:ROUT9 data is valid when LOCKN is low, otherwise ROUT0:ROUT9 is invalid.

Pins ROUT0: ROUT9 use pin RCLK as the reference to data. The polarity of the RCLK edge is controlled by the RCKR/ FN input [\(see Figure 28 on page 16\).](#page-15-1) ROUT0:ROUT9, LOCKN and RCLK outputs will drive a maximum of three CMOS Datasheet - Detailed Description

Resynchronization

When the deserializer PLL locks to the embedded clock edge, the deserializer LOCKN pin asserts a low. If the deserializer loses lock, pin LOCKN output will go high and the outputs (including RCLK) will enter tri-state.

The user's system monitors the pin LOCKN to detect a loss of synchronization. Upon detection, the system can arrange to pulse the serializer SYNC1 or SYNC2 pin to resynchronize.

Multiple resynchronization approaches are possible. It is mandatory to provide a feedback loop using pin LOCKN to control the SYNC request of the serializer (SYNC1 or SYNC2). Two SYNC pins are provided for multiple control in a multi-drop application. Sending SYNCPATs for resynchronization is desirable when lock times within a specific time are critical.

Powerdown

The low-power powerdown mode can be used while no data transfer is taking place. The serializer and deserializer use the powerdown mode to reduce power consumption by:

- The deserializer enters powerdown when pins PWDNN and REN are low.
- The serializer enters powerdown when pin PWDNN is driven low.

In powerdown, the PLL stops and the outputs enter tri-state, which disables load current and reduces supply current to the µA range.

Note: To exit powerdown, drive pin PWDNN high.

Before valid data exchanges between the serializer and deserializer, the devices must re-initialized and resynchronized to each other. Initialization of the serializer takes a maximum of 400 TCLK cycles. The deserializer will initialize and assert LOCKN high until lock to the Bus LVDS clock occurs.

Tri-State

The serializer enters tri-state when pin DEN is driven low. This puts both driver output pins (DO+ and DO-) into tristate. When DEN is driven high, the serializer returns to the previous state, as long as all other control pins remain static (SYNC1, SYNC2, PWDNN, TCKR/FN).

When pin REN is driven low, the deserializer enters tri-state. Consequently, the receiver output pins (ROUT0:ROUT9) and RCLK will enter tri-state. The LOCKN output remains active, reflecting the state of the PLL.

11 Application Information

Power Considerations

An all CMOS design of the serializer and deserializer makes them inherently low power devices. In addition, the constant current source nature of the bus LVDS outputs minimizes the slope of the speed vs. IDD curve of conventional CMOS designs.

The pins AVDD and DVDD should be bypassed with a 100nF and a 1nF ceramic capacitor in parallel. The 1nF capacitor should be closest to the pin.

Powering up the Deserializer

The AS1161 can be powered up at any time by following a proper sequence. The REFCLK input can be running before the deserializer powers up and it must be running in order for the deserializer to lock to incoming data. The deserializer outputs will remain in tri-state until the deserializer detects data transmission at its inputs and locks to the incoming data stream.

Inputs		Outputs				
PWDNN	REN	LOCKN	ROUTX	RCLK		
High	High	High (not locked)				
High	High	Low (not locked)	Active	Active		
Low	X (dont care)					
High	Low	Active ²				

Table 11. Deserializer Truth Table

1. Active indicates the RCLK will be running if the deserializer is locked. The timing of RCLK with respect to ROUT0:ROUT9 is determined by RCKR/FN. ROUT0:ROUT9 and RCLK are tri-stated when LOCKN is asserted high.

2. Active indicates the LOCKN output will reflect the state of the deserializer with regard to the selected data stream.

Transmitting Data

Once the serializer (AS1160) and deserializer (AS1161) are powered up, they must be phase locked to each other to transmit data. Phase locking occurs when the deserializer locks to incoming data or when the serializer sends patterns.

The serializer sends SYNCPATs whenever the SYNC1 or SYNC2 inputs are high. The LOCKN output of the deserializer remains high until it has locked to the incoming data stream. Connecting the LOCKN output of the deserializer to one of the SYNC inputs of the serializer will guarantee that enough SYNCPATs are sent to achieve deserializer lock [\(see Figure 36 on page 23\)](#page-22-0).

As long as the deserializer LOCKN output is low, valid data is presented at the deserializer outputs (ROUT0:ROUT9), except for the specific case of loss of lock during transmission [\(see Lock Loss Recovery on page 23\)](#page-22-1).

Figure 36. Typical Application

Noise Margin

The deserializer (AS1161) noise margin is the amount of input jitter (phase noise) that the deserializer can tolerate and still reliably receive data. Various environmental and systematic factors include:

- Serializer: TCLK jitter, VDD noise (noise bandwidth and out-of-band noise)
- Media: ISI (Inter Symbolic Interference), Large VCM shifts
- Deserializer: VDD noise

Lock Loss Recovery

In the case where the deserializer (AS1161) loses lock during data transmission, up to 3 cycles of data that were previously received can be invalid. This is due to the delay in the lock detection circuit. The lock detect circuit requires that invalid clock information be received 4 times in a row to indicate loss of lock. Since clock information has been lost, it is possible that data was also lost during these cycles. Therefore, after the deserializer re-locks to the incoming data stream and the deserializer LOCKN pin goes low, at least three previous data cycles should be suspect for bit errors.

The deserializer can re-lock to the incoming data stream by making the serializer re-send SYNCPATs.

Hot Insertion

As all BLVDS devices the AS1161 is hot pluggable but you have to follow some rules.

Hot insertion should be performed with pins making contact in the following order:

- Ground pins
- VDD pins
- I/O pins

Note: When removing the device, the pin groups should be removed in reverse order from insertion.

PCB Considerations

The serializer and deserializer should be placed as close to the PCB edge connector as possible. In multiple deserializer applications, the distance from the deserializer to the slot connector appears as a stub to the serializer driving the backplane traces. Longer stubs lower the impedance of the bus, increase the load on the serializer and lower the threshold margin at the deserializers. Deserializer devices should be placed much less than one inch from slot connectors. Because transition times are very fast on the serializer bus LVDS outputs, reducing stub lengths as much as possible is the best method to ensure signal integrity.

For bus LVDS applications the LVTTL, LVCMOS and bus LVDS signals should be separated from each other to prevent coupling into the bus lines. This can be achieved by using a four-layer PCB where the power, ground and input/ output signals are separated.

Transmission Media

The transmission line characteristics affect the performance of the AS1160/AS1161. It's recommended to use controlled-impedance media and to terminate at both ends of the transmission line [\(see Figure 37\).](#page-23-0) Twisted pair cables should be used due to their superior signal quality and the less EMI generation. Noise which is picked up as common mode in the twisted pair is rejected by the differential receiver.

It's important to eliminate reflections and to run the differential traces as close together as possible to ensure that the noise is coupled as common mode. Also take care of matching the electrical length of the traces to prevent a degradation of the magnetic field cancellation. To avoid an external magnetic field, the differential output signals should also be placed as close together as possible.

The potential of offsetting the ground levels of the serializer vs. the deserializer must be considered. The bus LVDS provides a +1.2V common mode range at the receiver inputs.

Figure 37. Double-Terminated Point-to-Point

The serializer/deserializer chipset can be used in many different topologies. Such as multidrop configurations [\(see Fig](#page-19-0)[ure 35 on page 20\),](#page-19-0) through a PCB trace or through twisted pair cable [\(see Figure 37\).](#page-23-0)

In point-to-point configurations, it's possible to terminated the transmission line only once at the receiver end. With only one termination the reflections and the differential signal swing are larger compared to a double termination.

Failsafe Biasing

The AS1161 has an input threshold sensitivity of ±75mV, which allows a greater differential noise margin.

However, in cases where the receiver input is not being actively driven, the increased sensitivity of the AS1161 can pickup noise as a signal and cause unintentional locking (e.g., when the input cable is left floating).

Figure 38. Failsafe Biasing Setup

External resistors can be added to the receiver circuit to prevent noise pickup as shown in [Figure 38](#page-24-0). In such circuits, the non-inverting receiver input is pulled up and the inverting receiver input is pulled down by high value resistors. The pull-up and pull-down resistors (R1 and R2 in [Figure 38](#page-24-0)) provide a current path through the termination resistor (RLOAD) which biases the receiver inputs when they are not connected to an active driver.

Note: The value of the pull-up and pull-down resistors should be chosen so that sufficient current is drawn to provide a +15mV drop across the termination resistor.

Signal Integrity

For a validation of the signal qualitiy in an application or in a simulation, the parameters tDJIT and tRNM can be used to generate an eye pattern mask.

tDJIT measures the transmitter's ability to place data bits in the ideal position to be sampled by the receiver. The typical tDJIT parameter of 50ps @ 66MHz indicates that the crossing point of the Tx data is 50ps before the ideal crossing point. The tDJITMIN and tDJITMAX parameters specify the earliest and latest time that a crossing will occur relative to the ideal position.

Figure 39. Eye Pattern Mask Generation and Signal Quality Validation

First of all, tRNM is calculated by measuring how much of the bit the receiver needs to ensure correct sampling. This calculated amount is subtracted from the ideal bit and what's left of it is available for external sources of noise and is called tRNM. It is the offset from tDJIT for the test mask within the eye opening.

The vertical limits of the mask are determined by the AS1161 receiver input threshold of ±75mV.

JTAG Test Modes

Instructions supported by the AS1160/AS1161 and its respective operational binary codes are shown in [Table 12.](#page-25-0)

Note: Boundary Scan Description Language (BSDL) model files for the AS1160 and the AS1161 are available on the internet.

Table 12. Instruction Codes

SAMPLE/PRELOAD

This is a mandatory instruction for the IEEE 1149.1 specification that supports two functions. The digital I/Os of the device can be sampled at the boundary scan test data register without interfering with the normal operation of the device. SAMPLE/PRELOAD also allows the device to shift data into the boundary scantest data register through TDI.

BYPASS

When the BYPASS instruction is latched into the instruction register, TDI connects to TDO through the 1-bit bypass test data register. This allows data to pass from TDI to TDO without affecting the device's normal operation.

EXTEST

Implemented at LVDS levels as a go/no-go test (e.g. missing cables).

IDCODE

Table 12. 22bit ID Code

The AS1160/AS1161 ID code is provided to the TDO output.

RUNBIST

An at-system-speed interconnect test instruction. It is executed in approximately 33mS (@ 66MHz system speed). There are two bits in the RX BIST data register for notification of PASS/FAIL and TEST_COMPLETE. Pass indicates

that the BER (bit-error-rate) is better than 10⁻⁷.

If both the AS1160 and the AS1161 have loaded the RUNBIST instruction into their instruction registers, both devices must move into the RTI state within 4K system clocks (at a SCLK of 66Mhz and TCK of 1MHz this allows for 66 TCK cycles). This is only an issue when both devices are not on the same scan chain or LSP, although, it can be a problem with some multi-drop devices.

Datasheet - Package Drawings and Markings

12 Package Drawings and Markings

The device is available in an CTBGA 49-bumps package.

Figure 40. CTBGA 49-bumps Package

Datasheet - Ordering Information

13 Ordering Information

The devices are available as the standard products shown in [Table 14.](#page-27-0)

Note: All products are RoHS compliant and Pb-free.

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