

#### DUAL-BAND GSM850/PCS1900 TRANSMIT MODULE

Package Style: Module 6.63mmx5.24mmx1.0mm

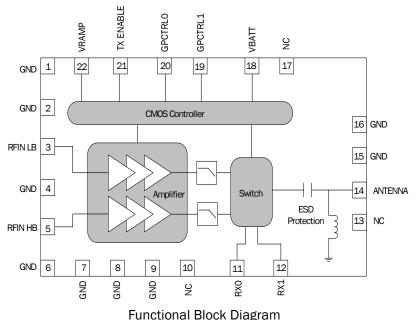


#### **Features**

- Enhanced Performance Transmit Module
- No External Routing
- High Efficiency at rated P<sub>OUT</sub>
   V<sub>BATT</sub>=3.5 V
   GSM850 42%
   PCS1900 38%
- Low RX Insertion Loss
- Symmetrical RX Ports
- OdBm to 6dBm Drive Level, >50dB of Dynamic Range
- Excellent ESD Protection at Antenna Port: 8kV
- Integrated Power Flattening Circuit
- V<sub>BATT</sub> Tracking Circuit

#### **Applications**

- 3V Dual-Band GSM/GPRS Handsets
- GSM850/PCS1900 Products
- GPRS Class 12 Compatible
- Portable Battery-Powered Equipment



## **Product Description**

The RF7167 is a dual band (GSM850/PCS1900) GSM/GPRS Class 12-compliant transmit module with two symmetrical receive ports. This transmit module builds upon RFMD's leading power amplifier with PowerStar® integrated power control technology, pHEMT switch technology, and integrated transmit filtering for best-in-class harmonic performance. The results are high performance, reduced solution size, and ease of implementation simplifies transmitter design. The device is designed for use as the final portion of the transmitter section in a GSM850/PCS1900 handset and eliminates the need for PA-to-antenna switch module matching network. The device provides  $50\Omega$  matched input and output ports, thereby requiring no external matching components.

The RF7167 features RFMD's latest integrated power-flattening circuit, which significantly reduces current and power variation into load mismatch. Additionally, a  $V_{BATT}$  tracking feature is incorporated to maintain switching perfromance as supply voltage decreases. The RF7167 also integrates an ESD filter to provide excellent ESD protection at the antenna port. The RF7167 is designed to provide maximum efficiency at rated  $P_{OUT}$ .

RF7167 Dual-Band GSM850/PCS1900 Transmit Module

RF7167SB Transmit Module 5-Piece Sample Pack RF7167PCBA-41X Fully Assembled Evaluation Board

# Optimum Technology Matching® Applied

☑ GaAs HBT	☐ SiGe BiCMOS	☑ GaAs pHEMT	☐ GaN HEMT
☐ GaAs MESFET	☐ Si BiCMOS	▼ Si CMOS	☐ RF MEMS
☐ InGaP HBT	☐ SiGe HBT	☐ Si BJT	☐ LDMOS

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#### **Absolute Maximum Ratings**

Parameter	Rating	Unit
Supply Voltage	-0.3 to +6.0	V
Power Control Voltage (V <sub>RAMP</sub> )	-0.3 to +1.8	V
Input RF Power	+10	dBm
Max Duty Cycle	50	%
Output Load VSWR	20:1	
Operating Case Temperature	-20 to +85	°C
Storage Temperature	-55 to +150	°C



#### Caution! ESD sensitive device.

Exceeding any one or a combination of the Absolute Maximum Rating conditions may cause permanent damage to the device. Extended application of Absolute Maximum Rating conditions to the device may reduce device reliability. Specified typical performance or functional operation of the device under Absolute Maximum Rating conditions is not implied.

RoHS status based on EUDirective 2002/95/EC (at time of this document revision).

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Parameter		Specification		Unit	Condition	
Farameter	Min.	Тур.	Max.	Ullit	Condition	
ESD						
ESD RF Ports			1000	V	HBM, JESD22-A114	
			1000	V	CDM, JEDECJESD22-C101	
ESD Antenna Port			8	kV	IEC 61000-4-2	
ESD Any Other Port			1000	V	HBM, JESD22-A114	
			1000	V	CDM, JEDEC JESD22-C101	
Overall Power Control V <sub>RAMP</sub>						
Power Control "ON"			1.8	V	Max. P <sub>OUT</sub>	
Power Control "OFF"		0.25		V	Min. P <sub>OUT</sub>	
V <sub>RAMP</sub> Input Capacitance		15	20	pF	DC to 200 kHz	
V <sub>RAMP</sub> Input Current			10	μА	V <sub>RAMP</sub> =V <sub>RAMP MAX</sub>	
Power Control Range		50		dB	V <sub>RAMP</sub> =0.25V to V <sub>RAMP MAX</sub>	
Overall Power Supply						
Power Supply Voltage	3.0	3.5	4.8	V	Operating Limits	
Power Supply Current		1	20	μА	P <sub>IN</sub> <-30dBm, TX Enable=Low, V <sub>RAMP</sub> =0.25V, Temp=-20°C to +85°C, V <sub>BATT</sub> =4.8V	
Overall Control Signals						
GpCtrl0, GpCtrl1 "Low"	0	0	0.5	V		
GpCtrl0, GpCtrl1 "High"	1.25	2.0	3.0	V		
GpCtrl0, GpCtrl1 "High Current"		1	2	uA		
TX Enable "Low"	0	0	0.5	V		
TX Enable "High"	1.25	2.0	3.0	V		
TX Enable "High Current"		1	2	uA		
RF Port Input and Output Imped- ance		50		Ω		

TX ENABLE	GpCtrl1	GpCtrl0	TX Module Mode
0	0	0	Low Power Mode (Stand-by)
0	1	0	RX 0
0	1	1	RX 1
1	1	0	GSM850 TX Mode
1	1	1	PCS1900 TX Mode





Baramatar		Specification	l	Unit	Condition
Parameter	Min.	Тур.	Max.	Unit	Condition
GSM850 Band					Nominal conditions unless otherwise stated. All unused ports are terminated.  VBATT=3.5V, PIN=3dBm, Temp=+25°C, TX Enable=High, VRAMP=1.8V TX Mode: GpCtrl1=High, GpCtrl0=Low, Duty Cycle=25%, Pulse Width=1154µs
Operating Frequency Range	824		849	MHz	
Input Power	0	3	6	dBm	Full P <sub>OUT</sub> guaranteed at minimum drive level.
Input VSWR			2.5:1		Over P <sub>OUT</sub> range (5 dBm to 33 dBm)
Maximum Output Power	33	33.7		dBm	Duty Cycle=25%, Pulse Width=1154μs
	30.5			dBm	$V_{BATT}$ =3.0V to 4.8V, $P_{IN}$ =0dBm to 6dBm, Temp=-20°C to +85°C, Duty Cycle=50%, Pulse Width=2308μs, $V_{RAMP}$ ≤1.8V
Minimum Power Into 3:1 VSWR	30			dBm	The measured delivered output power to the load with the mismatch loss already taken into account with 1dB variation margin.
Efficiency	36	42		%	Set V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33dBm
2nd Harmonic		-40*	-33	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33 dBm. *Typical value measured from worst case harmonic frequency across the band.
3rd Harmonic		-40*	-33	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33dBm. *Typical value measured from worst case harmonic frequency across the band.
All other harmonics up to 12.75 GHz			-33	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33dBm, also over all power levels (5dBm to 33dBm)
Non-Harmonic Spurious up to 12.75 GHz			-36	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33dBm, also over all power levels (5dBm to 33dBm)
Forward Isolation 1		-56	-41	dBm	TX Enable=Low, P <sub>IN</sub> =6dBm, V <sub>RAMP</sub> =0.25V
Forward Isolation 2		-28	-15	dBm	TX Enable=High, P <sub>IN</sub> =6dBm, V <sub>RAMP</sub> =0.25V
Output Noise Power					
869MHz to 894MHz		-88	-82	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33dBm,
1930MHz to 1990MHz		-118	-74	dBm	RBW=100 kHz
Output Load VSWR Stability (Spurious Emissions)			-36	dBm	VSWR=12:1; all phase angles (Set $V_{RAMP}$ = $V_{RAMP}$ rated for $P_{OUT}$ ≤33dBm into 50 $\Omega$ load; load switched to VSWR=12:1)
Output Load VSWR Ruggedness		damage or perma egradation to dev			VSWR=20:1; all phase angles (Set $V_{RAMP}$ = $V_{RAMP}$ rated for $P_{OUT}$ =33dBm into 50 $\Omega$ load; load switched to VSWR=20:1)



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Parameter	Min.	Тур.	Max.	Unit	Condition
PCS1900 Band					Nominal conditions unless otherwise stated. All unused ports are terminated.  VBATT=3.5V, PIN=3dBm, Temp=+25°C, TX Enable=High, VRAMP=1.8V  TX Mode: GpCtrl1=High, GpCtrl0=High, Duty Cycle=25%, Pulse Width=1154µs
Operating Frequency Range	1850		1910	MHz	
Input Power	0	3	6	dBm	Full P <sub>OUT</sub> guaranteed at minimum drive level.
Input VSWR			2.5:1		Over P <sub>OUT</sub> range (0dBm to 30dBm)
Maximum Output Power	30.0	31.5		dBm	Duty Cycle=25%, Pulse Width=1154μs
	28			dBm	$V_{BATT}$ =3.0V to 4.8V, $P_{IN}$ =0dBm to 6dBm, Temp=-20°C to +85°C, Duty Cycle=50%, Pulse Width=2308 μs, $V_{RAMP}$ ≤1.8V
Minimum Power Into 3:1 VSWR	27			dBm	The measured delivered output power to the load with the mismatch loss already taken into account with 1dB variation margin.
Efficiency	32	38		%	Set V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30dBm
2nd Harmonic		-39*	-33	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30 dBm. *Typical value measured from worst case harmonic frequency across the band.
3rd Harmonic		-40*	-33	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30 dBm. *Typical value measured from worst case harmonic frequency across the band.
All other harmonics up to 12.75 GHz			-33	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30dBm, also over all power levels (0dBm to 30dBm)
Non-Harmonic Spurious up to 12.75 GHz			-36	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30dBm, also over all power levels (0dBm to 30dBm)
Forward Isolation 1		-55	-53	dBm	TX Enable=Low, P <sub>IN</sub> =6dBm, V <sub>RAMP</sub> =0.25V
Forward Isolation 2		-25	-15	dBm	TX Enable=High, P <sub>IN</sub> =6dBm, V <sub>RAMP</sub> =0.25V
Output Noise Power					
869MHz to 894MHz		-103	-82	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30dBm, RBW=100kHz
1930MHz to 1990MHz		-84	-74	dBm	V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30dBm, RBW=100kHz
Output Load VSWR Stability (Spurious Emissions)			-36	dBm	VSWR=12:1; all phase angles (Set $V_{RAMP}$ = $V_{RAMP}$ rated for $P_{OUT}$ ≤30dBm into 50 $\Omega$ load; load switched to VSWR=12:1)
Output Load VSWR Ruggedness		damage or perma egradation to de			VSWR=20:1; all phase angles (Set $V_{RAMP} = V_{RAMP}$ rated for $P_{OUT} = 30  dBm$ into $50  \Omega$ load; load switched to VSWR=20:1)



Parameter		Specification		Unit	Condition
raiailletei	Min.	Тур.	Max.	Ullit	Condition
RX Section					Nominal conditions unless otherwise stated.  VBATT=3.5V, PIN=3dBm, Temp=+25°C,  TX Enable=Low, VRAMP=1.8V  RXO mode: GpCtrl1=High, GpCtrl0=Low  RX1 mode: GpCtrl1=High, GpCtrl0=High, RXO  Freq=869MHz to 894MHz,  RX1 Freq=1930MHz to 1990MHz
Insertion Loss GSM850 ANT-RX0/ RX1		1.1	1.3	dB	RXO Freq=869 MHz to 894 MHz
In-Band Ripple GSM850 ANT-RX0/RX1		0.2		dB	RXO Freq=869MHz to 894MHz
Input VSWR GSM850 ANT-RX0/RX1		1.5:1			RXO Freq=869 MHz to 894 MHz
Insertion Loss PCS1900 ANT-RX0/RX1		1.3	1.6	dB	Freq=1930MHz to 1990MHz
In-Band Ripple PCS1900 ANT-RX0/RX1		0.2		dB	Freq=1930MHz to 1990MHz
Input VSWR PCS1900 ANT-RX0/RX1		1.8:1			Freq=1930MHz to 1990MHz
TX Section					
Switch Leakage P <sub>OUT</sub> at RX Port GSM850 ANT-RXO/RX1		1	8	dBm	GSM850 TX mode: Freq=824MHz to 849MHz, GpCtrl1=High, GpCtrl0=Low, V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =33dBm at antenna port
Switch Leakage P <sub>OUT</sub> at RX Port PCS1900 ANT-RX0/RX1		5	6	dBm	PCS1900 TX mode: Freq=1850MHz to 1910MHz, GpCtrl1=High, GpCtrl0=High, V <sub>RAMP</sub> =V <sub>RAMP</sub> rated for P <sub>OUT</sub> =30dBm at antenna port

#### Notes:

Calculation example using typical values:  $P_{OUT}$  at Antenna -  $P_{OUT}$  at RX Port. Isolation LB=33-1=32dB, HB=30-5=25dB.

<sup>1.</sup> The insertion loss values listed are measured into  $50\Omega$  without matching. Improved performance can be obtained by properly matching the antenna/receiver ports.

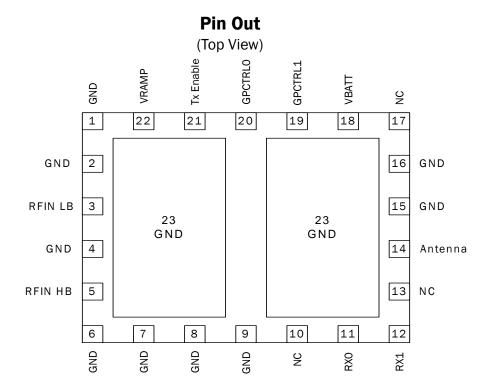
<sup>2.</sup> Isolation specification set to ensure at least the following isolation at rated power:



Pin	Function	Description	Interface Schematic
1	GND		
2	GND		
3	RFIN LB	RF input to the GSM850 band. This is a $50\Omega$ input.	LB RF IN O
4	GND		
5	RFIN HB	RF input to the PCS1900 band. This is a $50\Omega$ input.	HB RF IN O
6	GND		
7	GND		
8	GND		
9	GND		
10	NC		
11	RX0	RXO port of antenna switch. This is a $50\Omega$ output. RXO is interchangeable with RX1.	RX850
12	RX1	RX1 port of antenna switch. This is a $50\Omega$ output. RX1 is interchangeable with RX0.	RX1900
13	NC		
14	ANT	Antenna port of the switch. This is a $50\Omega$ output. Provides DC blocking as well as ESD protection.	
15	GND		
16	GND		
17	NC		
18	VBATT	Power supply for the module. This should be connected to the battery terminal using as wide a trace as possible.	
19	GPCTRL1	Control pin that together with CtrlO selects band of operation.	
20	GPCTRL0	Control pin that together with Ctrl1 selects band of operation.	
21	TX ENABLE	This signal enables the PA module for operation with a logic high. The switch is put in TX mode determined by GpCtrlO and GpCtrl1.	TVENABLE O
22	VRAMP	V <sub>RAMP</sub> ramping signal from DAC. A simple RC filter is integrated into the RF7167 module. V <sub>RAMP</sub> may or may not require additional filtering depending on the baseband selected.	TX ENABLE O TX ON
23	GND		
		I .	









## **Theory of Operation**

#### **Product Description**

The RF7167 is a dual-band, transmit module (TXM) with fully-integrated power control functionality, harmonic filtering, band selectivity, and TX/RX switching. The TXM is self-contained, with  $50\Omega$  I/O terminals with two symmetrical RX ports allowing dual-band operation. The power control function eliminates all power control circuitry, including directional couplers, diode detectors, and power control ASICs, etc. The power control capability provides  $50\,\mathrm{dB}$  continuous control range and  $70\,\mathrm{dB}$  of total control range, using a DAC-compatible, analog voltage input. The TX Enable feature provides for PA activation (TX mode) or RX mode/standby. Internal switching provides a low-loss, low-distortion path from the antenna port to the TX path (or RX port), while maintaining proper isolation. Integrated filtering provides ETSI-compliant harmonic suppression at the antenna port even under high mismatch conditions, which is important as modern antennas often present a load that significantly deviates from nominal impedance.

#### Overview

The RF7167 simplifies the phone design by eliminating the need for the complicated control loop, harmonic filters, and TX/RX switch along with their associated matching components. The power control loop can be driven directly from the DAC output in the baseband circuit. The module has two RX ports for GSM850 and PCS1900 bands of operation. The two RX ports are symmetrical; they can be used either for GSM850 or PCS1900. To control the mode of operation, there are three logic control signals: TX Enable, GpCtrl0, and GpCtrl1. The RF7167 offers high efficiency at the rated P<sub>OUT</sub> as backed-off efficiency is improved in this TXM.

#### **Power Ramping and Timing**

The RF7167 should be powered on according to the power-on sequence below. It is designed to prevent operation of the amplifier under conditions that could cause damage to the device or erratic operation.

There are some setup times associated with the control signals of the RF7167. The most important of these is the settling time between TXEN going high and when  $V_{RAMP}$  can begin to increase. This time is often referred to as the "pedestal" and is required so that the internal power control loop and bias circuitry can settle after being turned on. The RF7167 requires at least  $2\mu s$  or two quarter-bit times for proper settling of the power control loop.

The power-down sequence is in opposite order of the power-on sequence. As described in the figure below,  $V_{BATT}$  is applied first to provide bias to the silicon control chip. Then the RF drive is applied. Finally, when TXEN is high, The  $V_{RAMP}$  signal is held at a constant 0.25V, and  $2\,\mu s$  later,  $V_{RAMP}$  begins to ramp up. The shape of  $V_{RAMP}$  is important for maintaining the switching transients. The basic shape of the ramping function should be raised cosine to achieve best transient performance.



#### **Power On Sequence**

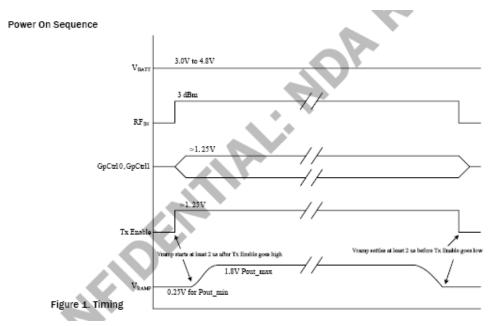


Figure 1. Timing Diagram

- 1. Apply V<sub>BATT</sub>
- 2. Apply GpCtrl0, GpCtrl1, RFIN and TX Enable
- 3. Apply V<sub>RAMP</sub> at least 2 µs after TX Enable
- 4. The Power Down Sequence is in opposite order of the Power On Sequence

#### Power Flattening and V<sub>BATT</sub> Tracking

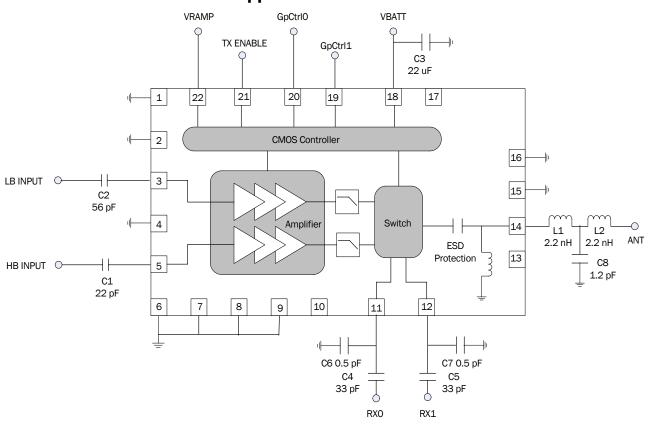
The RF7167 has an integrated power flattening circuit that reduces the amount of current variation when a mismatch is presented to the output of the PA. When a mismatch is presented to the output of the PA, its output impedance is varied and could present a load that will increase output power. As the output power increases, so does current consumption. The current consumption can become very high if not monitored and limited. The power flattening circuit is integrated onto the CMOS controller and requires no input from the user.

Into a mismatch, the current varies as the phase changes. The power flattening circuit monitors current through an internal sense resistor. As the current changes, the loop is adjusted in order to maintain current. The result is flatter power and reduced current into mismatch.

The RF7167 also incorporates a  $V_{BATT}$  tracking feature that eliminates the need for the transceiver/baseband to regulate the ramping signal as the supply voltage decreases. The internal circuit monitors the supply voltage and adjusts the ramping signal such that the switching spectrum is minimally impacted.



# **Application Schematic**



<sup>\*</sup>All inputs, outputs, and antenna traces are 50 W micro strip.

\*\*VBATT capacitor value may change depending on application.

\*\*\*RX0 and RX1 usually connect to SAW filters; C4 and C5 are used to block dc voltage present on the RX ports. C6 and C7 are used to match the RX ports to a 50 W filter.

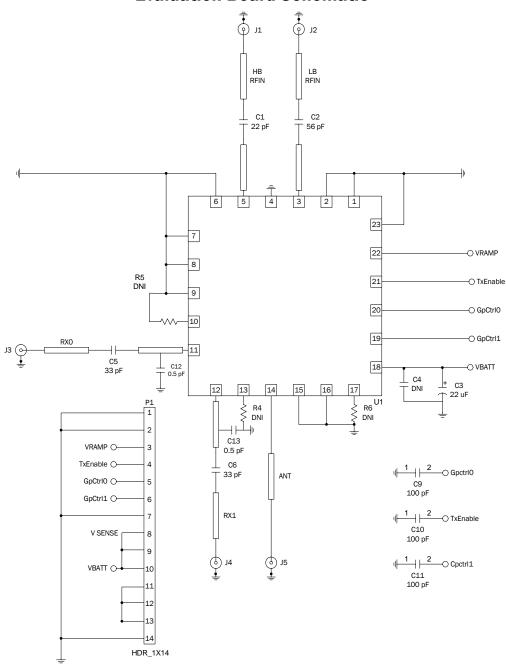
\*\*\*\*FX0 and RX1 usually connect to SAW filters; C4 and C5 are used to block dc voltage present on the RX ports. C6 and C7 are used to match the RX ports to a 50 W filter.

\*\*\*\*\*FX0 and RX1 usually connect to SAW filters; C4 and C5 are used to block dc voltage present on the RX ports. C6 and C7 are used to match the RX ports to a 50 W filter. amplifier.

\*\*\*\*\*L1, L2, and C8 is a suggested external LPF to suppress higher order harmonics. Actual component values will depend on the application.



## **Evaluation Board Schematic**

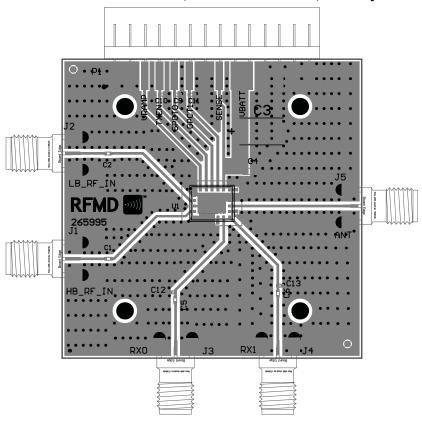


Notes: C9, C10, and C11 are optional decoupling capacitors which may not be needed in application. C5 and C6 are used to block the DC voltage present on the RX ports. C12 and C13 are used to match the RX ports to a  $50\Omega$  filter.



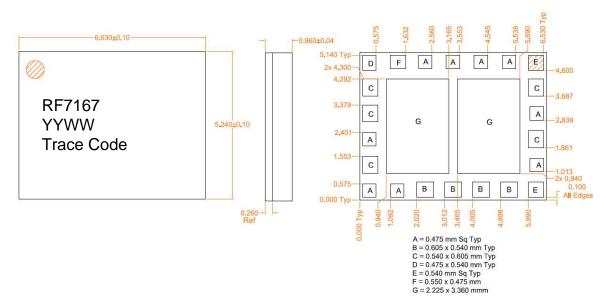
# Evaluation Board Layout Board Size 2.0" x 2.0"

Board Thickness 0.032", Board Material FR-4, Multi-layer





# **Package Drawing**



#### Notes:

YY indicates year, WW indicates work week, and Trace Code is a sequential number assigned at device assembly. Shaded areas represent Pin 1 location.



## **PCB Design Requirements**

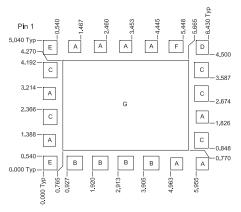
#### **PCB Surface Finish**

The PCB surface finish used for RFMD's qualification process is electroless nickel, immersion gold. Typical thickness is  $3\mu$ inch to  $8\mu$ inch gold over  $180\mu$ inch nickel.

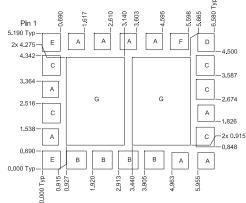
#### **PCB Land Pattern Recommendation**

PCB land patterns for RFMD components are based on IPC-7351 standards and RFMD empirical data. The pad pattern shown has been developed and tested for optimized assembly at RFMD. The PCB land pattern has been developed to accommodate lead and package tolerances. Since surface mount processes vary from company to company, careful process development is recommended.

#### **PCB Metal Land and Solder Mask Pattern**



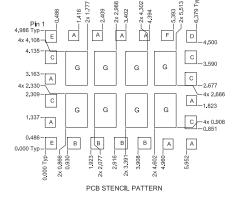
A = 0.475 mm Sq Typ
B = 0.605 x 0.540 mm Typ
C = 0.540 x 0.540 mm Typ
D = 0.475 x 0.540 mm Typ
E = 0.540 mm Sq Typ
F = 0.550 x 0.475 mm
G = 4.900 x 3.500 mm



A = 0.625 mm Sq Typ B = 0.755 x 0.690 mm Typ C = 0.690 x 0.755 mm Typ D = 0.625 x 0.690 mm Typ E = 0.690 mm Sq Typ F = 0.700 x 0.625 mm G = 2.225 x 3.360 mm Typ

PCB METAL LAND PATTERN

PCB SOLDER MASK PATTERN



A = 0.428 mm Sq Typ B = 0.544 x 0.486 mm Typ C = 0.486 x 0.544 mm Typ D = 0.428 x 0.486 mm Typ E = 0.486 mm Sq Typ F = 0.495 x 0.428 mm G = 0.911 x 1.422 mm Typ

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## **Tape and Reel**

Carrier tape basic dimensions are based on EIA 481. The pocket is designed to hold the part for shipping and loading onto SMT manufacturing equipment, while protecting the body and the solder terminals from damaging stresses. The individual pocket design can vary from vendor to vendor, but width and pitch will be consistent.

Carrier tape is wound or placed onto a shipping reel either 330mm (13 inches) in diameter or 178mm (7 inches) in diameter. The center hub design is large enough to ensure the radius formed by the carrier tape around it does not put unnecessary stress on the parts.

Prior to shipping, moisture sensitive parts (MSL level 2a-5a) are baked and placed into the pockets of the carrier tape. A cover tape is sealed over the top of the entire length of the carrier tape. The reel is sealed in a moisture barrier ESD bag with the appropriate units of desiccant and a humidity indicator card, which is placed in a cardboard shipping box. It is important to note that unused moisture sensitive parts need to be resealed in the moisture barrier bag. If the reels exceed the exposure limit and need to be rebaked, most carrier tape and shipping reels are not rated as bakeable at 125°C. If baking is required, devices may be baked according to section 4, table 4-1, of Joint Industry Standard IPC/JEDEC J-STD-033.

The table below provides information for carrier tape and reels used for shipping the devices described in this document.

#### Tape and Reel

RFMD Part Number	Reel Diameter Inch (mm)	Hub Diameter Inch (mm)	Width (mm)	Pocket Pitch (mm)	Feed	Units per Reel
RF7167TR13	13 (330)	4 (102)	12	8	Single	2500
RF7167TR7	7 (178)	2.4 (61)	12	8	Single	750

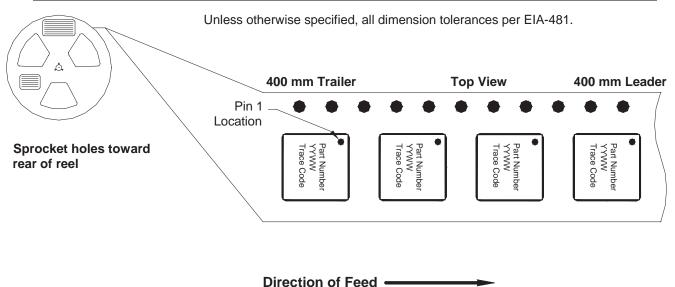


Figure 1. 5.24mmx6.63mm (Carrier Tape Drawing with Part Orientation)



#### **RoHS\* Banned Material Content**

RoHS Compliant: Yes
Package total weight in grams (g): 0.121
Compliance Date Code: Bill of Materials Revision: Pb Free Category: e4

Bill of Materials	Parts Per Million (PPM)						
bill of Materials	Pb	Cd	Hg	Cr VI	PBB	PBDE	
Die	0	0	0	0	0	0	
Molding Compound	0	0	0	0	0	0	
Lead Frame	0	0	0	0	0	0	
Die Attach Epoxy	0	0	0	0	0	0	
Wire	0	0	0	0	0	0	
Solder Plating	0	0	0	0	0	0	

This RoHS banned material content declaration was prepared solely on information, including analytical data, provided to RFMD by its suppliers, and applies to the Bill of Materials (BOM) revision noted above.

<sup>\*</sup> DIRECTIVE 2002/95/EC OF THE EUROPEAN PARLIAMENT AND OF THE COUNCIL of 27 January 2003 on the restriction of the use of certain hazardous substances in electrical and electronic equipment