

Low-cost, Simple 4-string LED Drivers with External Current Sink MOSFETs, 5000:1 Dimming Range and Per String PWM Input

Datasheet Brief

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General Description

The Atmel® LED Drivers-MSL2041 and MSL2042 compact, high-power LED string controllers use external current control MOSFETs to sink up to 1A per string, with string current matching of ±0.5%. The MSL2041/2 drive four parallel strings of LEDs and offer fault detection and management of open circuit and short circuit LEDs. The MSL2041 features four PWM inputs that allow independent frequency, dimming and phasing of each string, while the MSL2042 offers one PWM input for frequency and dimming control of all four strings, and automatically phase shifts the string drive signals. Peak string currents are set using current sense (FET source) resistors and adjustable with an internal 8-bit DAC.

The MSL2041/2 adaptively control up to two DC-DC converters that power the LED strings, using Atmel's Adaptive SourcePower™ technology. These Efficiency Optimizers minimize power use while maintaining LED current accuracy. Multiple MSL2041/2s cascade to automatically negotiate the optimum power supply voltage when driving more than four strings from a single power supply.

The MSL2041/2 features fault control for open-circuit strings, LED shortcircuits and device over-temperature conditions. When a string open-circuit or LED short-circuit condition is detected, the MSL2041/2 turn off the faulty string and pull the open-drain fault output low.

The MSL2041/2 feature stand-alone operation, and the basic circuit requires just one to four external PWM dimming inputs. An I²C serial interface is provided to allow optional control and monitoring of the various fault detection and Adaptive SourcePower parameters, but is not required for operation.

The MSL2041/2 are offered in the 32-pin, 300mil SOP package, and operate over the -40°C to +85°C temperature range.

Applications

- \cdot LCD-TVs
- PC Monitors
- Industrial Displays
- • General Illumination
- Street-lighting
- Post-regulated or Offline Powered LED Strings

Ordering Information

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Key Features

- Drives Four Parallel High Power LED Strings
- Up to 1A LED String Current with External N-channel MOSFETS
- • Operates Stand-alone, Basic Circuit Needs Only PWM Input(s)
- Four PWM Inputs Allow Individual Frequency, Brightness and Phase Control of each LED String (MSL2041)
- One PWM Input Controls the Frequency and Brightness of the Automatically Phase Shifted Strings (MSL2042)
- • 8-bit Adaptive SourcePower™ Correction Optimizes String Power Supply for Maximum Efficiency
- Multiple MSL2041/2s Share a String Supply and Automatically Negotiate the Optimum Supply Voltage
- ±0.5% Current Matching Between Strings
- String Open-circuit and LED Short-Circuit Fault Detection and Protection
- External MOSFETs Offer Flexibility of LEDs Used in Each String
- I²C Serial Interface Allows Optional Control of Device Functions and Faults
- • 32-pin 300mil SOP Package
- -40°C To +85°C Operating Temperature Range
- Lead-Free, Halogen-free, RoHS Compliant Package

Application Circuit

Package and Pinout - SOP

Package Dimensions: 32 Pin 20.52mm x 7.49mm x 2.49mm SOP (1.27mm pin pitch)

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Pin Descriptions

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Absolute Maximum Ratings

Voltage - With Respect to GND (SOP), EP/GND (TQFN)

Electrical Characteristics

VIN = 12V, $T_A = T_{MIN}$ to T_{MAX} , unless otherwise noted. Typical values are at VIN = 12V, $T_A = +25$ °C.

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Note 1. Minimum SCL clock frequency is limited by the bus timeout feature, which resets the serial bus interface if either SDA or SCL is held low for t TIMEOUT.

Note 2. t_{VDACK} = SCL LOW to SDA (out) LOW acknowledge time.

Note 3. t_{VDDAT} = minimum SDA output data-valid time following SCL LOW transition.

Note 4. A master device must internally provide an SDA hold time of at least 300ns to ensure an SCL low state.

Note 5. The maximum SDA and SCL rise times is 300ns. The maximum SDA fall time is 250ns. This allows series protection resistors to be connected between SDA and SCL inputs and the SDA/SCL bus lines without exceeding the maximum allowable rise time.

Note 6. MSL2041/2 includes input filters on SDA and SCL that suppress noise less than 50ns.

Note 7. Parameter is guaranteed by design and not production tested.

Block Diagram

Figure 1. Atmel LED Drivers-MSL2041/2 Block Diagram

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Typical Application Circuit

Figure 2. Atmel LED Drivers-MSL2042 driving four LED strings at 350mA peak current per string, controlling a single power supply

Detailed Description

The MSL2041 and MSL2042 are highly integrated, flexible, four-string LED drivers that use external MOSFETs to allow high string currents, and include power supply control to maximize efficiency of up to two external string power supplies. Optimized for stand-alone operation they require only external PWM signal(s), a few external components (including the string drive N-Channel MOSFETs) and an external string power supply. The MSL2041/2s four MOSFET gate drive outputs, G0 - G3, are optimized to drive FETs with a maximum gate voltage threshold of 3V.

The MSL2041/2 LED drivers provide simple control of LED brightness through both peak current and external PWM drive controls. Peak current control, set by external FET source resisters, offers excellent color consistency, while pulse width control allows simple brightness management. Multiple devices easily connect together to drive more than four LED strings while maintaining optimum system efficiency. An active low fault output activates when either a string open circuit or an LED short circuit condition is detected and verified. The MSL2041/2 are intended for stand-alone operation but offer additional string control and monitoring through a 1MHz I²C/SMBus compatible serial interface. Use of the serial interface is not required for operation.

The MSL2041 offers four PWM inputs that directly control the four string drive outputs, while the MSL2042 requires only a single PWM input signal and features automatic, progressive phase spreading of the four string drive signals. With phase spreading a ¼ PWM frame time delay is calculated and applied progressively to the string drive signals. Phase spreading helps reduce both the transient load on the LED power supply, and the power supply output capacitor size requirement.

The Adaptive SourcePower Efficiency Optimizer (EO) outputs control a wide range of different architectures of external DC/DC and AC/DC converters. Multiple drivers in a system communicate with each other in real time to select an optimized operating voltage for the LEDs. The EO allows design of the power supply for the worst case Forward Voltage (V_f) of the LEDs without worrying about excessive power dissipation issues, while ensuring that the LED drive system is operating at optimum

efficiency. During start-up the EO automatically reduces the string power supply voltage to the minimum value required to keep the LEDs in current regulation. The EO periodically performs re-optimization to compensate for changes of the LED's forward voltage, and to assure continued optimum power savings. Additionally, all string drivers are continually monitored for proper operation, and if any of the LED strings become starved for current the Efficiency Optimizer automatically increases the string power supply voltage to bring the string back in to current regulation.

Setting the Maximum LED String Current with the FET Source Resistor $R_{\rm c}$

The maximum string current, I_{EPI} for each string is set by a shunt resistor, $R_{S'}$ connected to ground from the source terminal of the string drive MOSFET (Figure 1, page 6). Determine the resistor value using

$$
R_s = \left(\frac{127}{255}\right) * \left(\frac{0.50196}{I_{LED}}\right) \Omega
$$

(where 127 is the default value of ISTR, String Current Control register 0x0E). For example, a full-scale LED current of 350mA returns R_s = 0.715 Ω (to the nearest 1% resistor value).

LED String Fault Response

The MSL2041/2 monitor the LED strings to detect LED short-circuit and string open-circuit faults (Figure 3). When verified, all string faults force the open drain fault output FLTB low.

After power-up, when shorted LEDs are verified in a string the string is disabled and no longer monitored by the Efficiency Optimizer. The short circuit threshold is 6V (typical) and the additive voltage drop lost from the shorted LEDs, plus the headroom required for the external FET, must be equal to or greater than the 6V threshold to generate a fault. Typically, two LEDs in a string must be shorted to cause a short circuit fault,

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but because LEDs differ, the number of shorted LEDs required to generate a fault varies. The current fold-back option, available through the serial interface, slightly changes the fault response when an LED short circuit event is suspected.

A string with an open circuit LED is off by default, and when this situation is verified the faulty string is disabled and no longer monitored by the Efficiency Optimizer.

Toggling EN low and then high clears all faults and the MSL2041/2 begin to control and monitor all strings as if experiencing an initial power-up. Fault conditions that persist re-establish fault responses.

Faulty strings are flagged in the fault registers. When using the serial interface, fault conditions are typically read in response to FLTB pulling low.

Figure 3. Open-circuit and Short-circuit Detection Block Diagram

Over Temperature Shutdown

The MSL2041/2 includes an automatic overtemperature shutdown. If the die temperature exceeds 135°C, the device turns off, just as if the enable input EN is forced low. When the die temperature drops below 120°C the device wakes up again and turns on as if experiencing an initial power-up.

Connecting the Efficiency Optimizer to an LED String Power Supply and Selecting Resistors

The MSL2041/2 are designed to control LED string power supplies that use a voltage divider $(R_{TOP}$ and R_{ROTOM} in Figure 4) to set output voltage, and whose regulation feedback voltage is not more than 3.5V - $\mathrm{V_{F}}$. The Efficiency Optimizer improves power efficiency by injecting a current of between 0µA and 280.5µA into the voltage divider of the external power supply, dynamically adjusting the power supply's output to the minimum voltage required by the LED strings.

Each of the two EOs monitors two LED strings. Strings zero and one are assigned to FBO1, and strings two and three are assigned to FBO2 (Table 1). When a single supply is used for all four strings connect FBO2 to FBI1 (Figure 4), as explained in the next section "*Using Multiple EOs/Devices to Control a Common Power Supply*". The MSL2041/2 then automatically maximizes efficiency for all strings. When two supplies are used, connect FBO1 to the supply powering strings zero and one, and connect FBO2 to the supply powering strings two and three (Figure 5). For clarity, Figure 4 and Figure 5 do not show the Source and Drain connections between the devices and the MOSFETs.

Table 1. String EO Assignments

Figure 4. EO Configuration When Using a Single String Power Supply

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Figure 5. EO Configuration When Using Two String Power Supplies

To select R_{TOP} and R_{BOTTOM} first determine $V_{\text{OUT(MIN)}}$ and $V_{\text{OUT}(\text{MAX})}$, the minimum and maximum string supply voltage limits, using:

$$
V_{OUT(MIN)} = (V_{f(MIN)} * [#ofLEDs]) + 0.5 ,
$$

and

$$
V_{OUT(MAX)} = (V_{f(MAX)} * [#ofLEDs]) + 0.5
$$
,

where $\mathsf{V}_{_{\mathsf{f}(\mathsf{MIN})}}$ and $\mathsf{V}_{_{\mathsf{f}(\mathsf{MAX})}}$ are the LED's minimum and maximum forward voltage drops at the peak current set by $\mathsf{R}_{\mathsf{S}^*}$ For example, if the LED data are $\mathsf{V}_{\mathsf{f}(\mathsf{MIN})}$ = 3.5V and $V_{f(MAX)} = 3.8V$, and ten LEDs are used in a string, then the total minimum and maximum voltage drop across a string is 35V and 38V. Adding an allowance of 0.5V for the string drive MOSFET headroom brings $V_{\text{OUT(MIN)}}$ to 35.5V and $V_{\text{OUT} (MAX)}$ to 38.5V. Then determine R_{TOP} using:

$$
R_{TOP} = \frac{V_{OUT(MAX)} - V_{OUT(MIN)}}{I_{FBOn(MAX)}},
$$

where $I_{FBOn(MAX)}$ is the 280.5µA maximum output current of the Efficiency Optimizer outputs FBOn (280.5µA = 1.1µA * 255, the current per LSB of the FBO DAC times the maximum DAC count). Finally, determine $R_{\scriptscriptstyle\rm BOTDM}$ using:

$$
R_{\text{BOTTOM}} = R_{\text{TOP}} * \frac{V_{\text{FB}}}{V_{\text{OUT}(\text{MAX})} - V_{\text{FB}}}
$$

where V_{FB} is the regulation feedback voltage of the power supply. Place a diode (1N4148 or similar) between FBOn and the supply's feedback node to protect the MSL2041/2 against current flow into FBOn.

Once configured, determine the change in power supply output voltage in response to a change in FBOn output current using:

$$
\Delta V_{OUT} = \Delta I_{FBOn} * R_{TOP}
$$

Assure that the power supply settling time for a voltage step size of 1.1 μ A $*$ R_{TOP} is less than the 4ms EO Step-hold duration time.

Using Multiple EOs/Devices to Control a Common Power Supply

Cascade multiple Efficiency Optimizers (EOs), either within the same device or across multiple devices, into a chain configuration (Figure 6), with the FBIn of one EO connected to the FBOn of the next. Connect the first FBOn to the power supply feedback resistor node through a diode (1N4148 or similar) placed close to the power supply feedback node, and unused FBIn inputs to ground as close to the MSL2041/2 as possible. The chained EOs work together to ensure that the system operates at optimum efficiency. Note that the accuracy of the feedback chain may degrade through each link of the FBIn/FBOn chain by 2% (typical). Derate the maximum FBOn current using:

 $I_{FBO(MAX/MIN)} = 280.5 \mu A^* (0.98)^{N-1}$,

where N is the number of EOs connected in series. Use I $I_{FBOn(MAX/MIN)}$ in the above R_{TOP} resistor equation for the term $I_{FBOn(MAX)}$ instead of using 280.5µA.

Take care in laying out the traces for the Efficiency Optimizer connections. Minimize the FBIn/FBOn trace lengths as much as possible. Do not route the signals close to traces with large variations in voltage or current, because noise may couple into FBIn. If these traces must be routed near noisy signals, shield them from noise by using ground planes or guard traces. For clarity, Figure 6 shows Source and Drain connections only for unused outputs 2 and 3 of device two. Note that because of the interplay between EOs and the automatic fault response behavior, when both strings monitored by a single EO fault and turn off, that the string supply is forced to its maximum value and all remaining active strings typically detect short circuit faults and also turn off.

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Figure 6. EO Chain Configuration of Two Devices, Six Strings and a Single String Power Supply

Choosing the Drain Resistor R_{D}

The drain resistor R_D connects the MSL2041/2 to the Drain of the external MOSFET. Choose $\mathsf{R}_{_\mathsf{D}}$ using:

$$
R_{D} = \left(\frac{V_{OUT(MAX)} - N(V_{F(DARK)}) - 22.5}{I_{DARK}}\right) - 1 \times 10^{5} \Omega.
$$

where $V_{\text{OUT(MAX)}}$ is the value calculated above in the section "*Connecting the Efficiency Optimizer to an LED String Power Supply and Selecting Resistors*" beginning on page 12, N is the number of LEDs in the string, I_{DARK} is the maximum allowable string off current and $V_{F(DARK)}$ is the LED forward voltage drop at I_{DARK} . When the value calculated for $\mathsf{R}_\texttt{D} < 0$ use 0 Ω .

LED manufacturers typically do not publish I_{DARK} and $V_{F(DARK)}$ information. One way to determine these numbers is to use the following method.

> Set up the test circuit of Figure 7. Adjust R1 until the current meter indicates I_{DARK} (choose I I_{DARK} < 1mA). Use a volt meter to measure the voltage at the anode of the LED (A), and then at the cathode of the LED (B). Subtract the voltage measured at B from that measured at A to determine $V_{F(DARK)}$. Some typical values determined using this method are listed in Table 2.

Figure 7. Test Circuit for Determining V_{F(DARK)}

Table 2. Some Typical I_{DARK} and V_{F(DARK)} Values Determined **Using Figure 7**

LED TYPE	LED PART#	$I_{DARK}(\mu A)$	$V_{F(DARK)}(V)$
LOW POWER	LW Y1SG	1.72	2.285
MEDIUM POWER	LW G6SP	1.67	2.276
HIGH POWER	LXLW-PWC1	1.72	2.195

Large values of R_D may cause false LED short circuit faults. Discharge of the parasitic capacitance at the Dn node through a large R_D holds the node above the string fault threshold for longer than the LED short circuit verification time. The addition of a feed-forward capacitor, C_{FF} in Figure 8, mitigates this issue. The value for C_{FF} depends upon the amount of parasitic capacitance at the Dn node and the size of $\mathsf{R}_{_{\mathsf{D}^{\prime}}}$ but $\mathsf{C}_{_{\mathsf{FF}}}$ = 15pF is an appropriate first approximation.

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Direct PWM Control of the LED Strings

An external PWM signal applied to the inputs PWM0 - PWM3 (MSL2041) or PWM0 (MSL2042) allows direct control over the strings frequency and duty cycle. The PWM inputs recognize signals of DC to 50kHz, and 0% to 100% duty cycle. The MSL2042, which allows only a single PWM input, calculates and applies a progressive delay of 1/4th the PWM frame successively to strings one - three, while string zero follows the PWM input directly.

Register Map Summary

Control the MSL2041/2 using the registers in the range 0x00 - 0x18. Register bit values always revert to their default values (Table 4) when EN is taken high. Do not write to registers not listed in Table 3.

Table 3. Register Map

*** Read Only Registers**

Register Power-up Defaults

Register power-up default values are shown in Table 4.

Table 4. Register Power-up Defaults

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