

**LOW DROPOUT CMOS VOLTAGE REGULATOR**[www.sii-ic.com](http://www.sii-ic.com)

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Rev.3.0\_00

The S-818 Series is a positive voltage regulator developed by CMOS technology and featured by low dropout voltage, high output voltage accuracy and low current consumption. Built-in low on-resistance transistor provides low dropout voltage and large output current. A ceramic capacitor of 2  $\mu$ F or more can be used as an output capacitor. A shutdown circuit ensures long battery life. The SOT-23-5 miniaturized package and the SOT-89-5 package are recommended for configuring portable devices and large output current applications, respectively.

**■ Features**

- Low current consumption: At operation mode: Typ. 30  $\mu$ A, Max. 40  $\mu$ A  
At shutdown mode: Typ. 100 nA, Max. 500 nA
- Output voltage: 2.0 to 6.0 V, selectable in 0.1 V steps.
- High accuracy output voltage:  $\pm 2.0\%$
- Peak output current: 200 mA capable (3.0 V output product,  $V_{IN}=4$  V)<sup>\*1</sup>  
300 mA capable (5.0 V output product,  $V_{IN}=6$  V)<sup>\*1</sup>
- Low dropout voltage: Typ. 170 mV (5.0 V output product,  $I_{OUT}=60$  mA)
- A ceramic capacitor (2  $\mu$ F or more) can be used as an output capacitor.
- Built-in shutdown circuit
- Lead-free, Sn 100%, halogen-free<sup>\*2</sup>

\*1. Attention should be paid to the power dissipation of the package when the output current is large.

\*2. Refer to "■ Product Name Structure" for details.

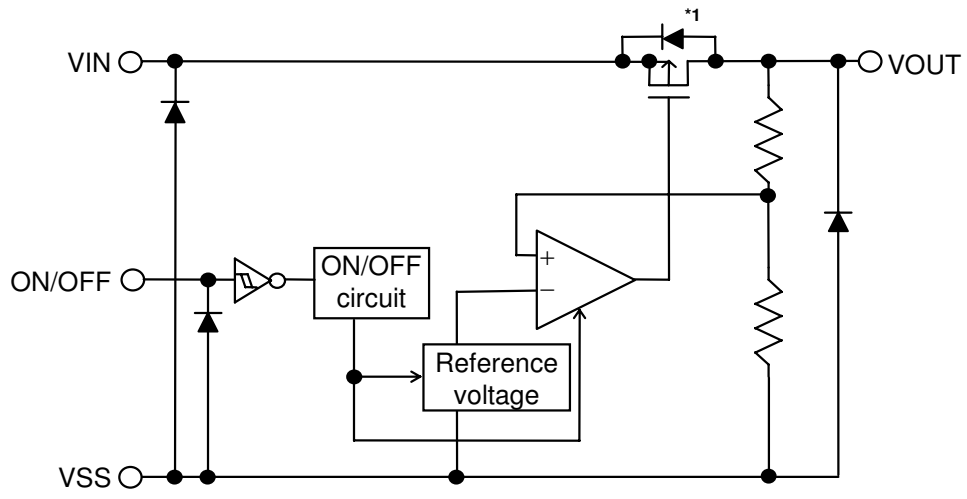
**■ Applications**

- Power source for battery-powered devices, personal communication devices and home electric/electronic appliances

**■ Packages**

- SOT-23-5
- SOT-89-5

■ Block Diagram

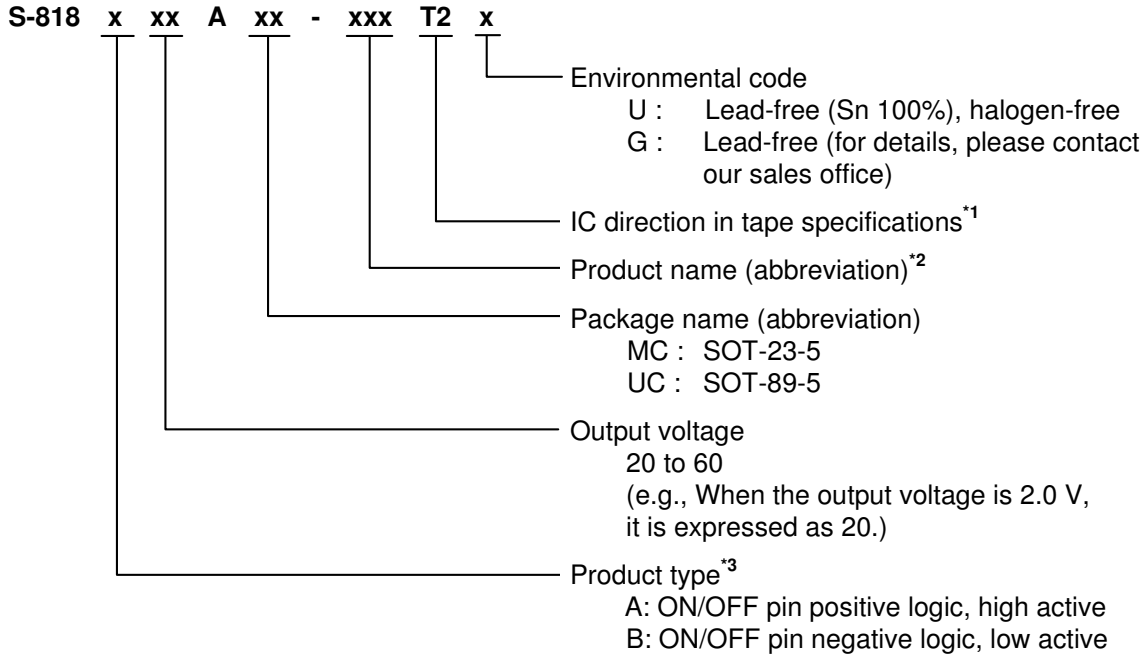


\*1. Parasitic diode

Figure 1

■ **Product Name Structure**

**1. Product Name**



- \*1. Refer to the tape specifications at the end of this book.
- \*2. Refer to the "Table 1" under the "3. Product name list".
- \*3. Refer to "3. ON/OFF pin (Shutdown pin)" in the "■ Operation".

**2. Package**

Package Name	Drawing Code		
	Package	Tape	Reel
SOT-23-5	MP005-A-P-SD	MP005-A-C-SD	MP005-A-R-SD
SOT-89-5	UP005-A-P-SD	UP005-A-C-SD	UP005-A-R-SD

**3. Product name list**

**Table 1**

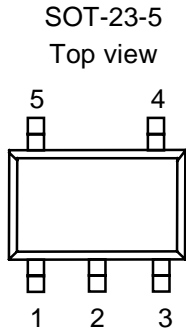
Output Voltage	SOT-23-5	SOT-89-5
2.0 V±2.0%	S-818A20AMC-BGAT2x	S-818A20AUC-BGAT2x
2.1 V±2.0%	S-818A21AMC-BGBT2x	S-818A21AUC-BGBT2x
2.2 V±2.0%	S-818A22AMC-BGCT2x	S-818A22AUC-BGCT2x
2.3 V±2.0%	S-818A23AMC-BGDT2x	S-818A23AUC-BGDT2x
2.4 V±2.0%	S-818A24AMC-BGET2x	S-818A24AUC-BGET2x
2.5 V±2.0%	S-818A25AMC-BGFT2x	S-818A25AUC-BGFT2x
2.6 V±2.0%	S-818A26AMC-BGGT2x	S-818A26AUC-BGGT2x
2.7 V±2.0%	S-818A27AMC-BGHT2x	S-818A27AUC-BGHT2x
2.8 V±2.0%	S-818A28AMC-BGIT2x	S-818A28AUC-BGIT2x
2.9 V±2.0%	S-818A29AMC-BGJT2x	S-818A29AUC-BGJT2x
3.0 V±2.0%	S-818A30AMC-BGKT2x	S-818A30AUC-BGKT2x
3.1 V±2.0%	S-818A31AMC-BGLT2x	S-818A31AUC-BGLT2x
3.2 V±2.0%	S-818A32AMC-BGMT2x	S-818A32AUC-BGMT2x
3.3 V±2.0%	S-818A33AMC-BGNT2x	S-818A33AUC-BGNT2x
3.4 V±2.0%	S-818A34AMC-BGOT2x	S-818A34AUC-BGOT2x
3.5 V±2.0%	S-818A35AMC-BGPT2x	S-818A35AUC-BGPT2x
3.6 V±2.0%	S-818A36AMC-BGQT2x	S-818A36AUC-BGQT2x
3.7 V±2.0%	S-818A37AMC-BGRT2x	S-818A37AUC-BGRT2x
3.8 V±2.0%	S-818A38AMC-BGST2x	S-818A38AUC-BGST2x
3.9 V±2.0%	S-818A39AMC-BGTT2x	S-818A39AUC-BGTT2x
4.0 V±2.0%	S-818A40AMC-BGUT2x	S-818A40AUC-BGUT2x
4.1 V±2.0%	S-818A41AMC-BGVT2x	S-818A41AUC-BGVT2x
4.2 V±2.0%	S-818A42AMC-BGWT2x	S-818A42AUC-BGWT2x
4.3 V±2.0%	S-818A43AMC-BGXT2x	S-818A43AUC-BGXT2x
4.4 V±2.0%	S-818A44AMC-BGYT2x	S-818A44AUC-BGYT2x
4.5 V±2.0%	S-818A45AMC-BGZT2x	S-818A45AUC-BGZT2x
4.6 V±2.0%	S-818A46AMC-BHAT2x	S-818A46AUC-BHAT2x
4.7 V±2.0%	S-818A47AMC-BHBT2x	S-818A47AUC-BHBT2x
4.8 V±2.0%	S-818A48AMC-BHCT2x	S-818A48AUC-BHCT2x
4.9 V±2.0%	S-818A49AMC-BHDT2x	S-818A49AUC-BHDT2x
5.0 V±2.0%	S-818A50AMC-BHET2x	S-818A50AUC-BHET2x
5.1 V±2.0%	S-818A51AMC-BHFT2x	S-818A51AUC-BHFT2x
5.2 V±2.0%	S-818A52AMC-BHGT2x	S-818A52AUC-BHGT2x
5.3 V±2.0%	S-818A53AMC-BHHT2x	S-818A53AUC-BHHT2x
5.4 V±2.0%	S-818A54AMC-BHIT2x	S-818A54AUC-BHIT2x
5.5 V±2.0%	S-818A55AMC-BHJT2x	S-818A55AUC-BHJT2x
5.6 V±2.0%	S-818A56AMC-BHKT2x	S-818A56AUC-BHKT2x
5.7 V±2.0%	S-818A57AMC-BHLT2x	S-818A57AUC-BHLT2x
5.8 V±2.0%	S-818A58AMC-BHMT2x	S-818A58AUC-BHMT2x
5.9 V±2.0%	S-818A59AMC-BHNT2x	S-818A59AUC-BHNT2x
6.0 V±2.0%	S-818A60AMC-BHOT2x	S-818A60AUC-BHOT2x

**Remark 1.** Please contact our sales office for type B products.

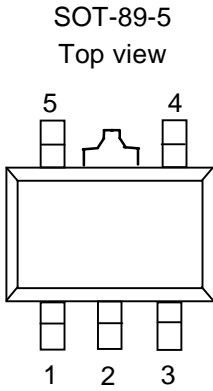
**2.** x: G or U

**3.** Please select products of environmental code = U for Sn 100%, halogen-free products.

■ **Pin Configurations**



**Figure 2**



**Figure 3**

**Table 2**

Pin No.	Symbol	Pin description
1	VIN	Input voltage pin
2	VSS	GND pin
3	ON/OFF	Shutdown pin
4	NC <sup>*1</sup>	No connection
5	VOUT	Output voltage pin

\*1. The NC pin is electrically open.  
The NC pin can be connected to VIN or VSS.

**Table 3**

Pin No.	Symbol	Pin description
1	VOUT	Output voltage pin
2	VSS	GND pin
3	NC <sup>*1</sup>	No connection
4	ON/OFF	Shutdown pin
5	VIN	Input voltage pin

\*1. The NC pin is electrically open.  
The NC pin can be connected to VIN or VSS.



### ■ Electrical Characteristics

**Table 5**

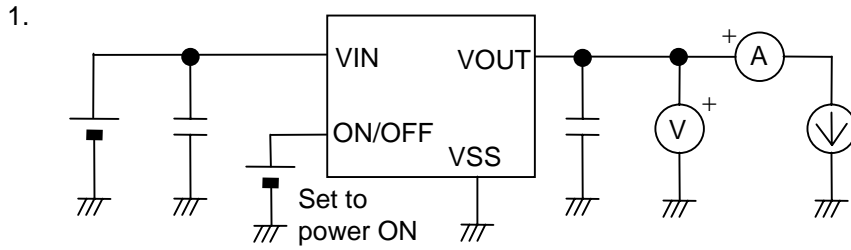
(Ta=25°C unless otherwise specified)

Parameter	Symbol	Conditions	Min.	Typ.	Max.	Unit	Test circuit	
Output voltage <sup>*1</sup>	$V_{OUT(E)}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $I_{OUT}=30\text{ mA}$	$V_{OUT(S)} \times 0.98$	$V_{OUT(S)}$	$V_{OUT(S)} \times 1.02$	V	1	
Output current <sup>*2</sup>	$I_{OUT}$	$V_{OUT(S)}+1\text{ V} \leq V_{IN} \leq 10\text{ V}$	$2.0\text{ V} \leq V_{OUT(S)} \leq 2.4\text{ V}$	$100^{*5}$	—	—	mA	3
			$2.5\text{ V} \leq V_{OUT(S)} \leq 2.9\text{ V}$	$150^{*5}$	—	—	mA	3
			$3.0\text{ V} \leq V_{OUT(S)} \leq 3.9\text{ V}$	$200^{*5}$	—	—	mA	3
			$4.0\text{ V} \leq V_{OUT(S)} \leq 4.9\text{ V}$	$250^{*5}$	—	—	mA	3
			$5.0\text{ V} \leq V_{OUT(S)} \leq 6.0\text{ V}$	$300^{*5}$	—	—	mA	3
Dropout voltage <sup>*3</sup>	$V_{drop}$	$I_{OUT}=60\text{ mA}$	$2.0\text{ V} \leq V_{OUT(S)} \leq 2.4\text{ V}$	—	0.51	0.87	V	1
			$2.5\text{ V} \leq V_{OUT(S)} \leq 2.9\text{ V}$	—	0.38	0.61	V	1
			$3.0\text{ V} \leq V_{OUT(S)} \leq 3.4\text{ V}$	—	0.30	0.44	V	1
			$3.5\text{ V} \leq V_{OUT(S)} \leq 3.9\text{ V}$	—	0.24	0.33	V	1
			$4.0\text{ V} \leq V_{OUT(S)} \leq 4.4\text{ V}$	—	0.20	0.26	V	1
			$4.5\text{ V} \leq V_{OUT(S)} \leq 4.9\text{ V}$	—	0.18	0.22	V	1
			$5.0\text{ V} \leq V_{OUT(S)} \leq 5.4\text{ V}$	—	0.17	0.21	V	1
		$5.5\text{ V} \leq V_{OUT(S)} \leq 6.0\text{ V}$	—	0.17	0.20	V	1	
Line regulation 1	$\frac{\Delta V_{OUT1}}{\Delta V_{IN} \cdot V_{OUT}}$	$V_{OUT(S)}+0.5\text{ V} \leq V_{IN} \leq 10\text{ V}$ , $I_{OUT}=30\text{ mA}$	—	0.05	0.2	%/V	1	
Line regulation 2	$\frac{\Delta V_{OUT2}}{\Delta V_{IN} \cdot V_{OUT}}$	$V_{OUT(S)}+0.5\text{ V} \leq V_{IN} \leq 10\text{ V}$ , $I_{OUT}=10\text{ }\mu\text{A}$	—	0.05	0.2	%/V	1	
Load regulation	$\Delta V_{OUT3}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $10\text{ }\mu\text{A} \leq I_{OUT} \leq 80\text{ mA}$	—	30	50	mV	1	
Output voltage temperature coefficient <sup>*4</sup>	$\frac{\Delta V_{OUT}}{\Delta T_a \cdot V_{OUT}}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $I_{OUT}=30\text{ mA}$ , $-40^\circ\text{C} \leq T_a \leq 85^\circ\text{C}$	—	$\pm 100$	—	ppm/ °C	1	
Current consumption at operation	$I_{SS1}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , ON/OFF pin=ON, no load	—	30	40	$\mu\text{A}$	2	
Current consumption at shutdown	$I_{SS2}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , ON/OFF pin=OFF, no load	—	0.1	0.5	$\mu\text{A}$	2	
Input voltage	$V_{IN}$	—	—	—	10	V	1	
Shutdown pin input voltage "H"	$V_{SH}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $R_L=1\text{ k}\Omega$ , Judged by $V_{OUT}$ output level.	1.5	—	—	V	4	
Shutdown pin input voltage "L"	$V_{SL}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $R_L=1\text{ k}\Omega$ , Judged by $V_{OUT}$ output level.	—	—	0.3	V	4	
Shutdown pin input current "H"	$I_{SH}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $V_{ON/OFF}=7\text{ V}$	-0.1	—	0.1	$\mu\text{A}$	4	
Shutdown pin input current "L"	$I_{SL}$	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $V_{ON/OFF}=0\text{ V}$	-0.1	—	0.1	$\mu\text{A}$	4	
Ripple rejection	$ RR $	$V_{IN}=V_{OUT(S)}+1\text{ V}$ , $f=100\text{ Hz}$ , $\Delta V_{rip}=0.5\text{ V p-p}$ , $I_{OUT}=30\text{ mA}$	—	45	—	dB	5	

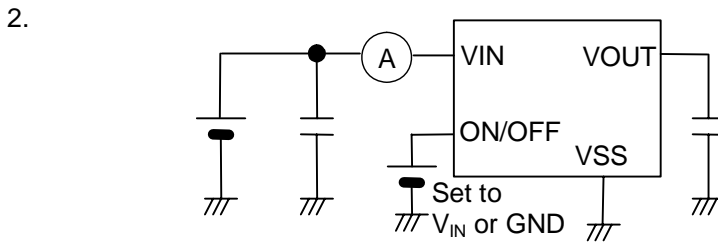
- \*1.  $V_{OUT(S)}$ =Specified output voltage  
 $V_{OUT(E)}$ =Effective output voltage  
i.e., The output voltage when fixing  $I_{OUT}$  (=30 mA) and inputting  $V_{OUT(S)}+1.0$  V.
- \*2. Output current at which output voltage becomes 95 % of  $V_{OUT(E)}$  after gradually increasing output current.
- \*3.  $V_{drop}=V_{IN1}^*1-(V_{OUT(E)}\times 0.98)$ 
  - \*1. The Input voltage at which output voltage becomes 98 % of  $V_{OUT(E)}$  after gradually decreasing input voltage.
- \*4. Output voltage shift by temperature [mV/°C] is calculated using the following equation.  
$$\frac{\Delta V_{OUT}}{\Delta T_a} [\text{mV}/^\circ\text{C}]^*1 = V_{OUT(S)}[\text{V}]^*2 \times \frac{\Delta V_{OUT}}{\Delta T_a \bullet V_{OUT}} [\text{ppm}/^\circ\text{C}]^*3 \div 1000$$
  - \*1. Temperature change ratio for output voltage
  - \*2. Specified output voltage
  - \*3. Output voltage temperature coefficient
- \*5. These figures mean that every part can supply output current at least to these values



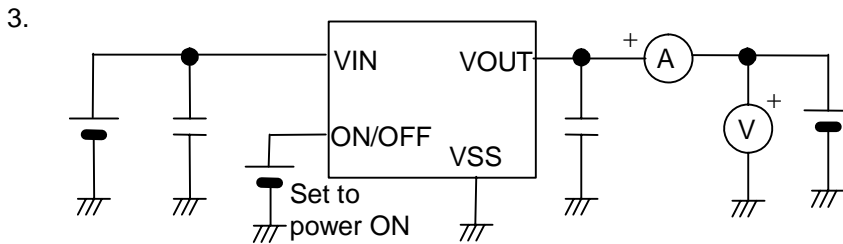
■ **Test Circuits**



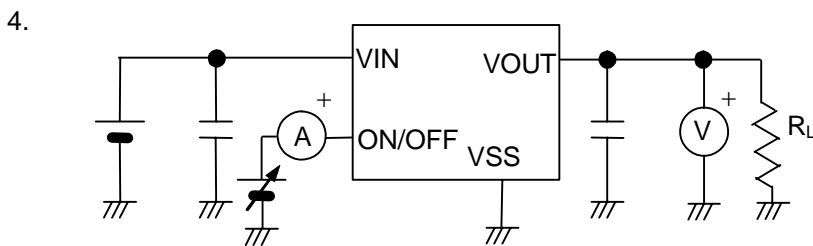
**Figure 5**



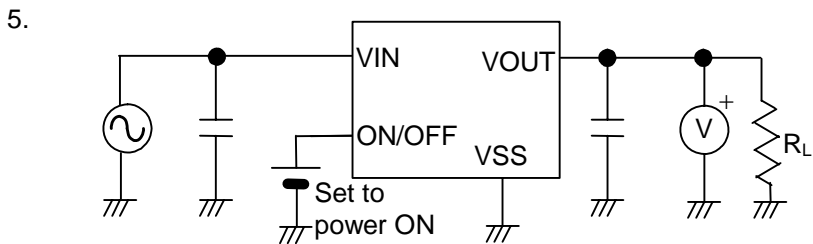
**Figure 6**



**Figure 7**



**Figure 8**



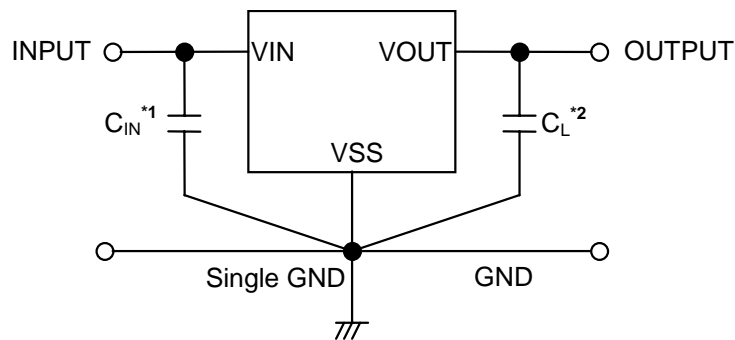
**Figure 9**

■ **Application Conditions**

Input capacitor ( $C_{IN}$ ):	0.47 $\mu$ F or more
Output capacitor ( $C_L$ ):	2 $\mu$ F or more
Equivalent series resistor (ESR):	10 $\Omega$ or less
Input series resistor ( $R_{IN}$ ):	10 $\Omega$ or less

**Caution** A general series regulator may oscillate, depending on the external components selected. Check that no oscillation occurs with the application using the above capacitor.

■ **Standard Circuit**



- \*1.  $C_{IN}$  is a capacitor used to stabilize input. Use a capacitor of 0.47  $\mu$ F or more
- \*2. In addition to a tantalum capacitor, a ceramic capacitor of 2.0  $\mu$ F or more can be used for  $C_L$ .

**Figure 10**

**Caution** The above connection diagram and constant will not guarantee successful operation. Perform through evaluation using the actual application to set the constant.

■ **Technical Terms**

**1. Low dropout voltage regulator**

The low dropout voltage regulator is a voltage regulator having a low dropout voltage characteristic due to the internal low on-resistance transistor.

**2. Output voltage ( $V_{OUT}$ )**

The accuracy of the output voltage is ensured at  $\pm 2.0\%$  under the specified conditions of input voltage, output current, and temperature, which differ product by product.

**Caution** When the above conditions are changed, the output voltage may vary and go out of the accuracy range of the output voltage. Refer to the “■ Electrical Characteristics” and “■ Characteristics” for details.

**3. Line regulation 1 ( $\Delta V_{OUT1}$ ) and Line regulation 2 ( $\Delta V_{OUT2}$ )**

Line regulation indicates the input voltage dependence of the output voltage. The value shows how much the output voltage changes due to the change of the input voltage when the output current is kept constant.

**4. Load regulation ( $\Delta V_{OUT3}$ )**

Load regulation indicates the output current dependence of output voltage. The value shows how much the output voltage changes due to the change of the output current when the input voltage is kept constant.

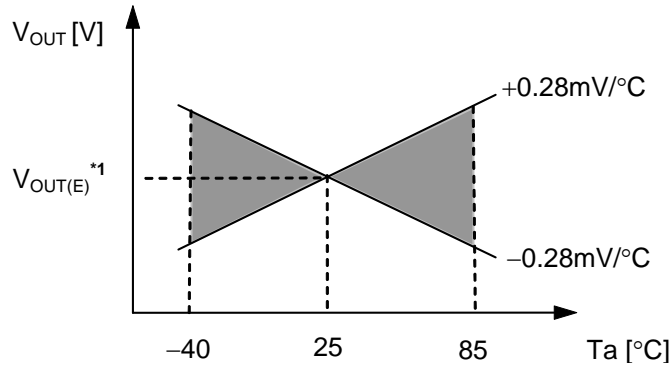
**5. Dropout voltage ( $V_{drop}$ )**

Let  $V_{IN1}$  be an input voltage where the output voltage falls to the 98 % of the actual output voltage ( $V_{OUT(E)}$ ) when gradually decreasing input voltage. The dropout voltage is the difference between the  $V_{IN1}$  and the resultant output voltage defined as following equation.

$$V_{drop} = V_{IN1} - (V_{OUT(E)} \times 0.98)$$

**6. Temperature coefficient of output voltage  $\left( \frac{\Delta V_{OUT}}{\Delta T_a \bullet V_{OUT}} \right)$**

The output voltage lies in the shaded area in the whole operating temperature shown in **Figure 11** when the temperature coefficient of the output voltage is  $\pm 100$  ppm/ $^{\circ}\text{C}$ .



\*1. The value of the output voltage measured at 25°C.

**Figure 11 Temperature coefficient of output voltage (Ex. Typ. product for S-818A28A)**

Temperature change ratio for output voltage [mV/ $^{\circ}\text{C}$ ] is calculated by using the following equation.

$$\frac{\Delta V_{OUT}}{\Delta T_a} [\text{mV}/^{\circ}\text{C}]^{*1} = V_{OUT(S)} [\text{V}]^{*2} \times \frac{\Delta V_{OUT}}{\Delta T_a \bullet V_{OUT}} [\text{ppm}/^{\circ}\text{C}]^{*3} \div 1000$$

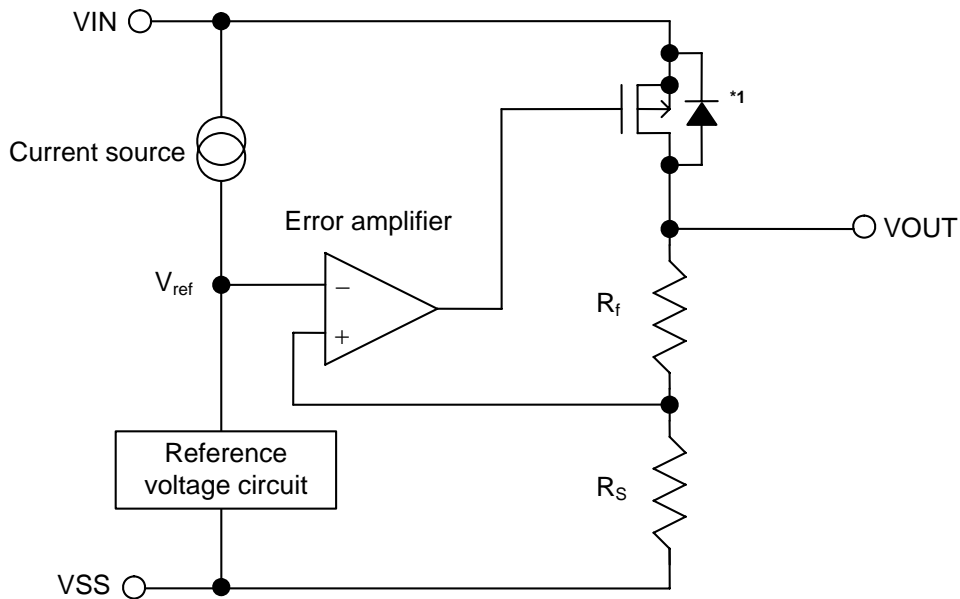
- \*1. Temperature change ratio for output voltage
- \*2. Specified output voltage
- \*3. Output voltage temperature coefficient

■ **Operation**

**1. Basic Operation**

**Figure 12** shows the block diagram of the S-818 Series.

The error amplifier compares a reference voltage ( $V_{ref}$ ) with the part of the output voltage divided by the feedback resistors  $R_s$  and  $R_f$ . It supplies the output transistor with the gate voltage, necessary to ensure certain output voltage free of any fluctuations of input voltage and temperature.



\*1. Parasitic diode

**Figure 12 Block diagram**

**2. Output Transistor**

The S-818 Series uses a Pch MOS FET as the output transistor.

Be sure that  $V_{OUT}$  does not exceed  $V_{IN}+0.3$  V to prevent the voltage regulator from being damaged due to inverse current flowing from VOUT pin through a parasitic diode to VIN pin.

### 3. ON/OFF pin (Shutdown pin)

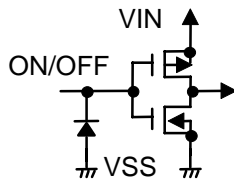
This pin activates and inactivates the regulator.

When the ON/OFF pin is switched to the shutdown level, the operation of all internal circuit stops, the built-in Pch MOS FET output transistor between VIN and VOUT pin is switched off, suppressing current consumption. The VOUT pin goes to the Vss level due to internal divided resistance of several MΩ between VOUT pin and VSS pin.

The structure of the ON/OFF pin is shown in **Figure 13**. Since the ON/OFF pin is neither pulled down nor pulled up internally, do not keep it in the floating state. Current consumption increases if a voltage of 0.3 V to  $V_{IN}-0.3$  V is applied to the ON/OFF pin. When the shutdown pin is not used, connect it to the VIN pin for product type "A" and to the VSS pin for product type "B".

**Table 6 ON/OFF pin function by product type**

Product type	ON/OFF pin	Internal circuit	VOUT pin voltage	Current consumption
A	"H": Power on	Operating	Set value	$I_{SS1}$
A	"L": Shutdown	Stop	Vss level	$I_{SS2}$
B	"H": Shutdown	Stop	Vss level	$I_{SS2}$
B	"L": Power on	Operating	Set value	$I_{SS1}$



**Figure 13 The structure of the ON/OFF Pin**

### ■ Selection of Output Capacitor ( $C_L$ )

The S-818 Series needs an output capacitor between VOUT pin and VSS pin for phase compensation. A small ceramic or an OS electrolyte capacitor of 2 μF or more can be used. When a tantalum or an aluminum electrolyte capacitor is used, the capacitance must be 2 μF or more and the ESR must be 10 Ω or less.

Attention should be paid not to cause an oscillation due to increase of ESR at low temperatures when an aluminum electrolyte capacitor is used.

Evaluate the performance including temperature characteristics before prototyping the circuit.

Overshoot and undershoot characteristics differ depending upon the type of the output capacitor. Refer to the " **$C_L$  dependence**" data in "**■ Transient Response Characteristics**".

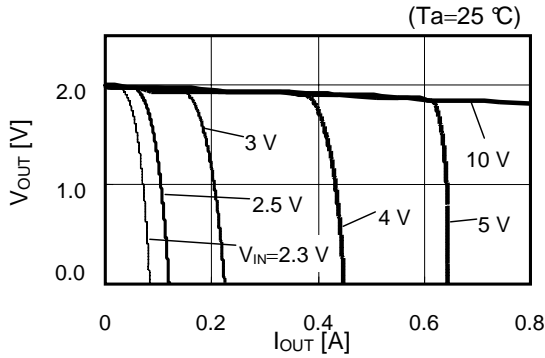
■ **Precautions**

- Wiring patterns for the VIN pin, VOUT pin and GND pin should be designed so that the impedance is low. When mounting an output capacitor ( $C_L$ ) or an input capacitor ( $C_{IN}$ ), the distance from the capacitor to the VOUT pin and to the VSS pin should be as short as possible.
- Note that output voltage may increase when a voltage regulator is used at low load current (Less than 10  $\mu$ A).
- To prevent oscillation, the external components should be used under the following conditions:
  - Input capacitor ( $C_{IN}$ ): 0.47 $\mu$ F or more
  - Output capacitor ( $C_L$ ): 2  $\mu$ F or more
  - Equivalent series resistance (ESR): 10  $\Omega$  or less
  - Input series resistance ( $R_{IN}$ ): 10  $\Omega$  or less
- The voltage regulator may oscillate when the impedance of the power supply is high and the input capacitor is small or not connected.
- The application condition for input voltage and load current should not exceed the package power dissipation.
- In determining output current, attention should be paid to the output current value specified and footnote \*5 in **Table 5** in the “■ **Electrical Characteristics**”.
- Do not apply an electrostatic discharge to this IC that exceeds the performance ratings of the built-in electrostatic protection circuit.
- SII claims no responsibility for any and all disputes arising out of or in connection with any infringement by products including this IC of patents owned by a third party.

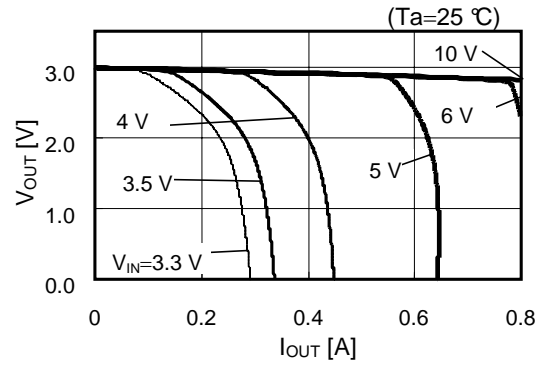
■ **Characteristics (Typical data)**

**1. Output Voltage ( $V_{OUT}$ ) vs. Output Current ( $I_{OUT}$ ) (When load current increases)**

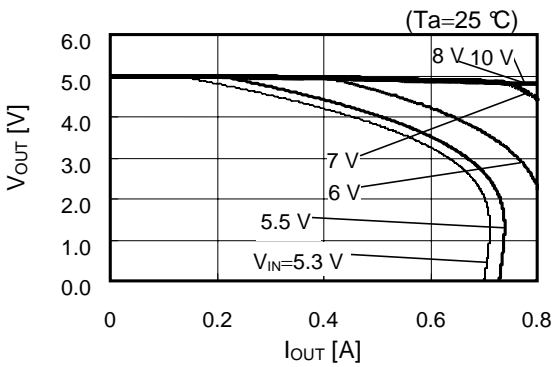
S-818A20A



S-818A30A



S-818A50A

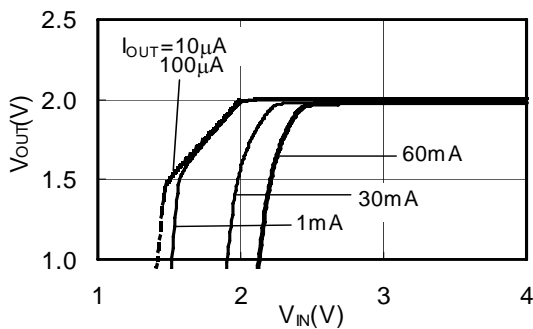


**Remark** In determining necessary output current, consider the following parameters:

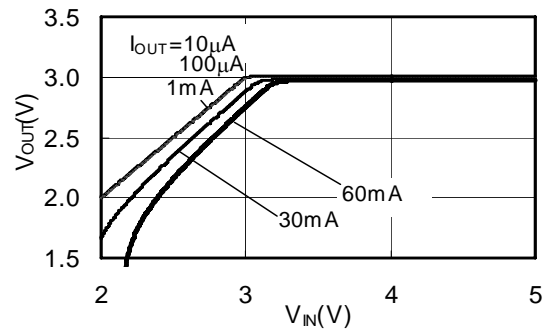
1. Output current value in the “■ **Electrical Characteristics**” and footnote \*5.
2. Power dissipation of the package

**2. Output voltage ( $V_{OUT}$ ) vs. Input voltage ( $V_{IN}$ )**

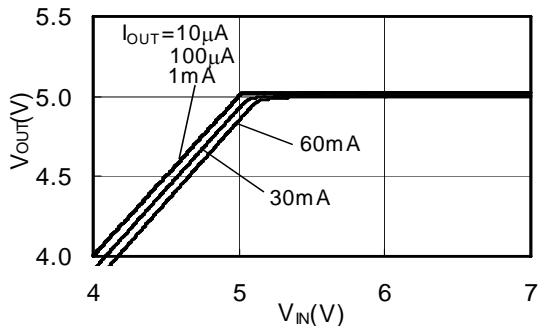
S-818A20A (Ta=25°C)



S-818A30A (Ta=25°C)

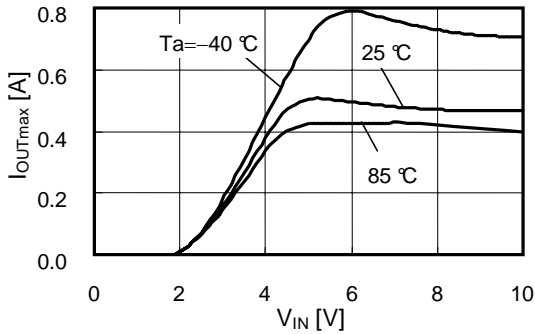


S-818A50A (Ta=25°C)

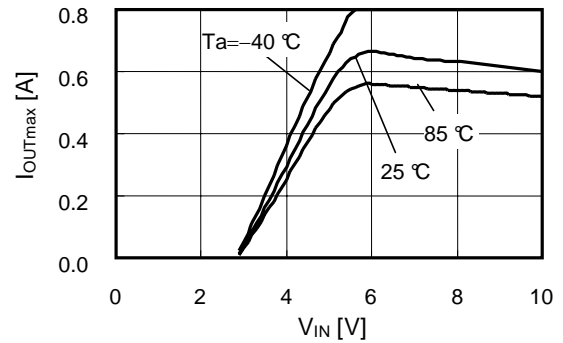


**3. Maximum output current ( $I_{OUTmax}$ ) vs. Input voltage ( $V_{IN}$ )**

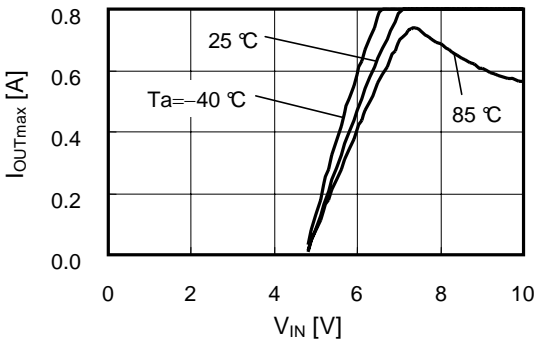
S-818A20A



S-818A30A



S-818A50A

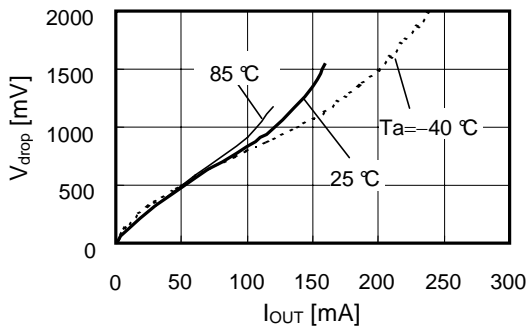


**Remark** In determining necessary output current, consider the following parameters:

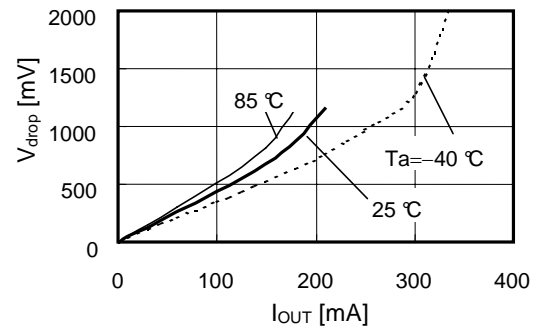
1. Output current value in the "■ **Electrical Characteristics**" and footnote \*5.
2. Power dissipation of the package

**4. Dropout voltage ( $V_{drop}$ ) vs. Output current ( $I_{OUT}$ )**

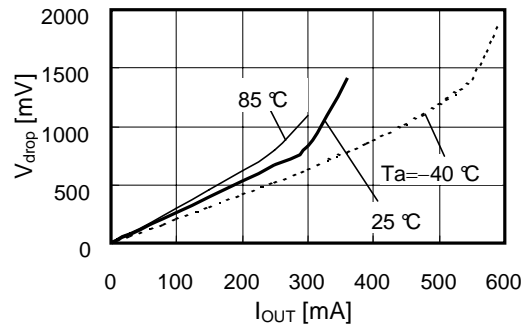
S-818A20A



S-818A30A



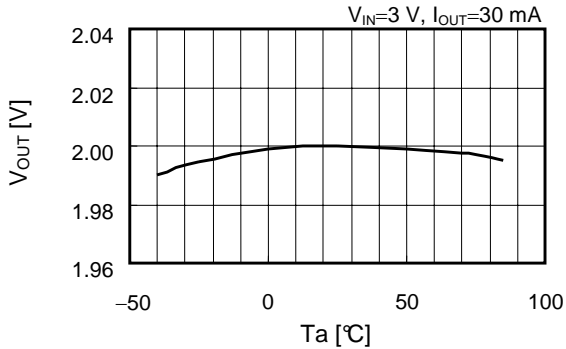
S-818A50A



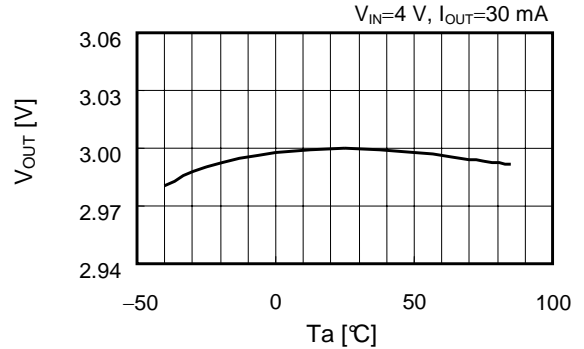


**5. Output voltage ( $V_{OUT}$ ) vs. Ambient temperature ( $T_a$ )**

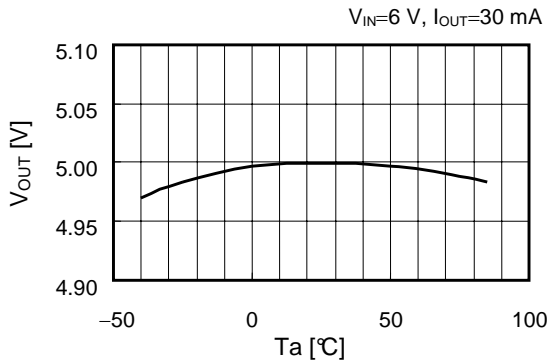
S-818A20A



S-818A30A

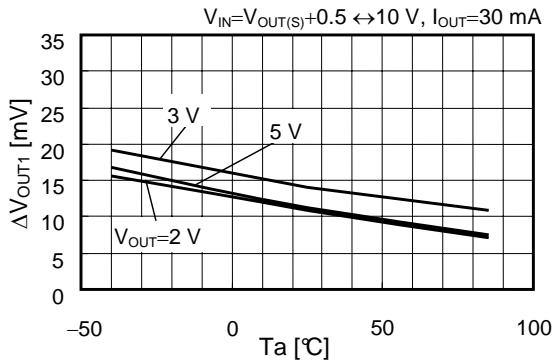


S-818A50A



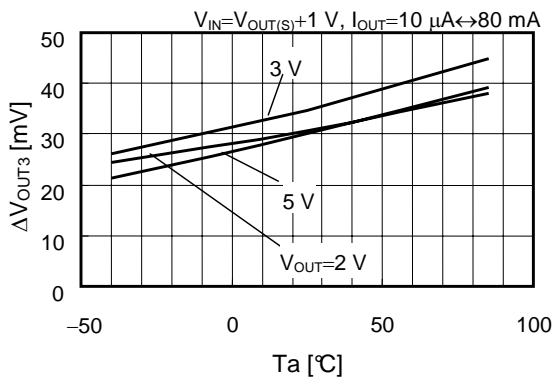
**6. Line regulation ( $\Delta V_{OUT1}$ ) vs. Ambient temperature ( $T_a$ )**

S-818A20A/S-818A30A/S-818A50A

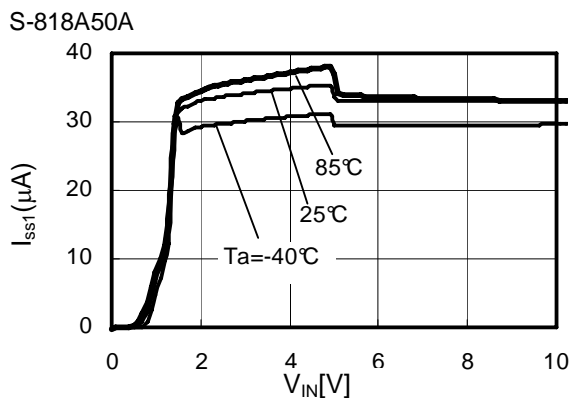
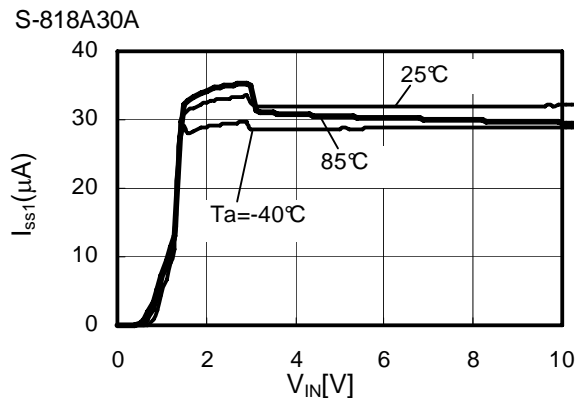
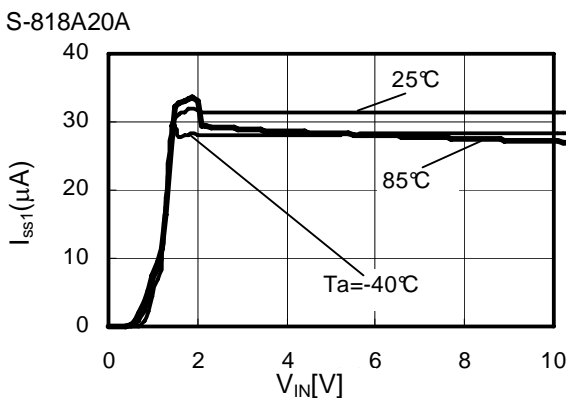


**7. Load regulation ( $\Delta V_{OUT3}$ ) vs. Ambient temperature ( $T_a$ )**

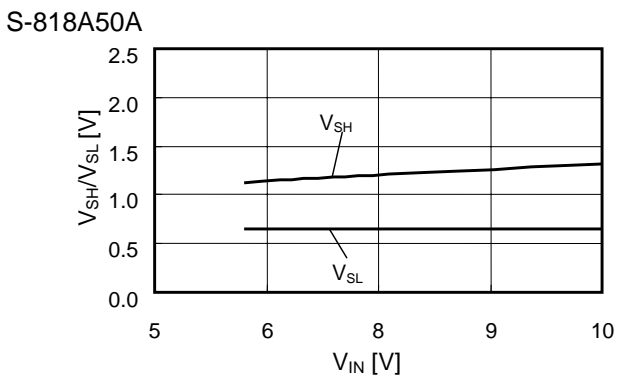
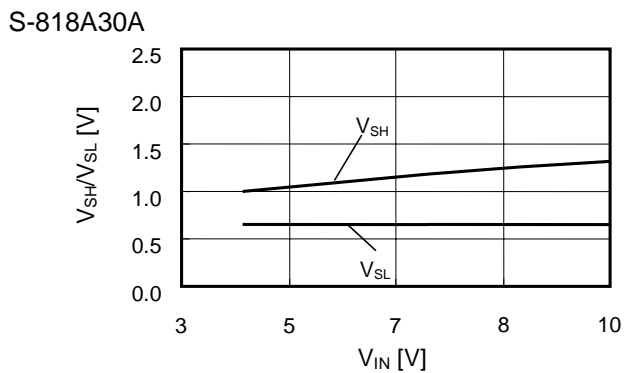
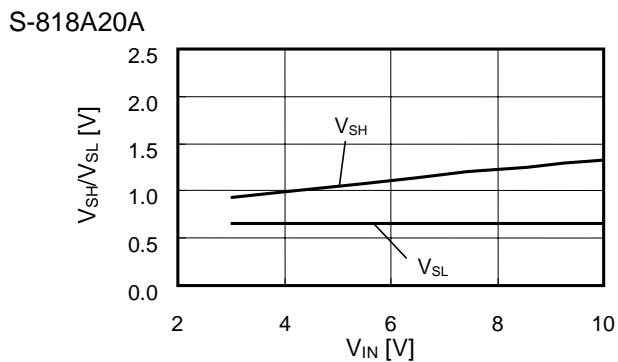
S-818A20A/S-818A30A/S-818A50A



**8. Current consumption ( $I_{SS1}$ ) vs. Input voltage ( $V_{IN}$ )**

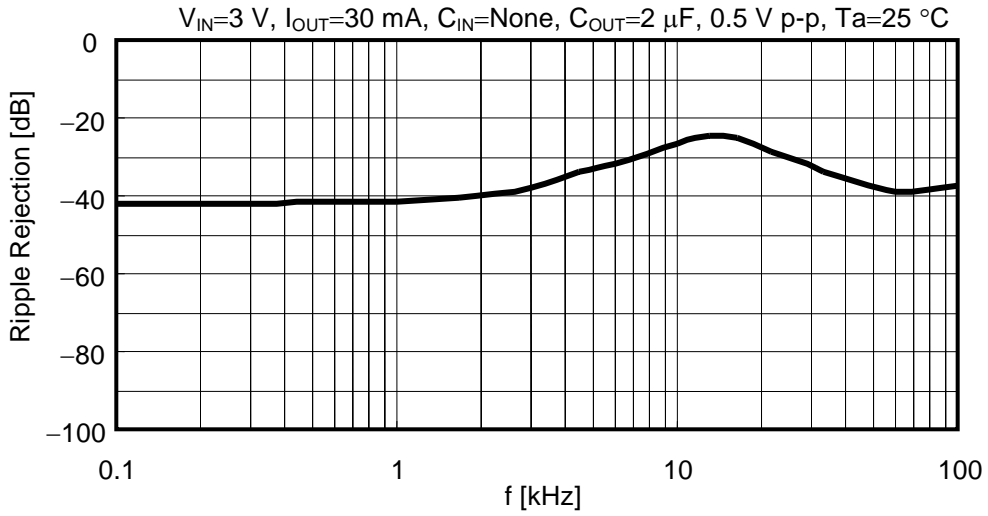


**9. Threshold voltage of ON/OFF pin ( $V_{SH}/V_{SL}$ ) vs. Input voltage ( $V_{IN}$ )**

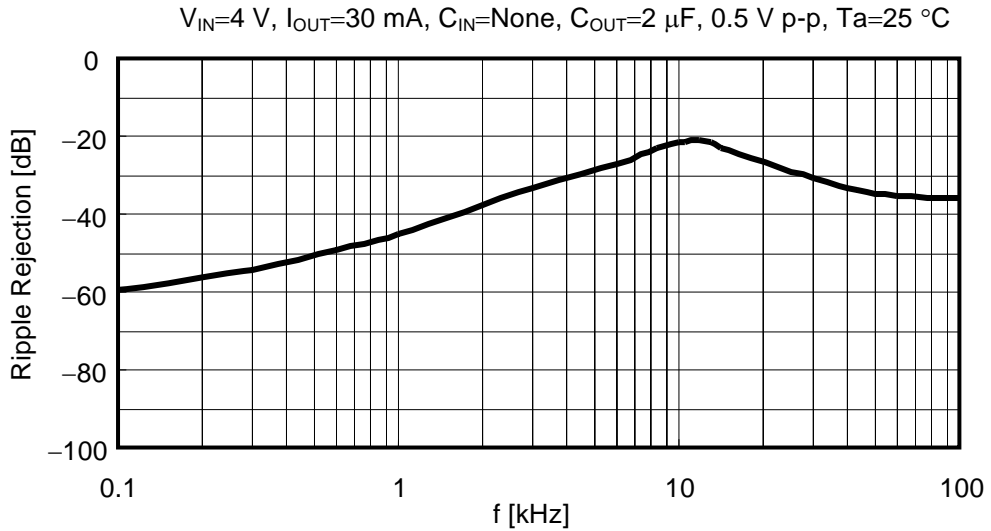


**10. Ripple rejection**

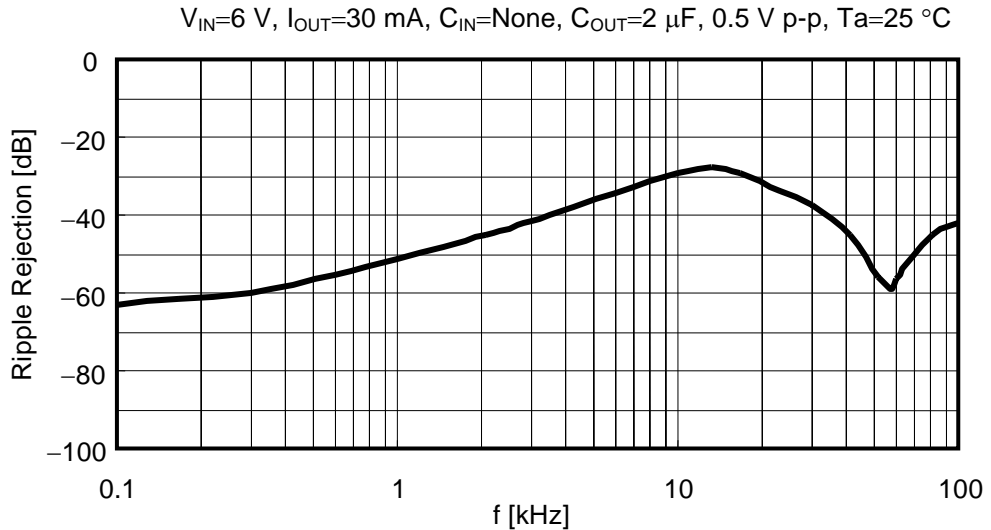
S-818A20A



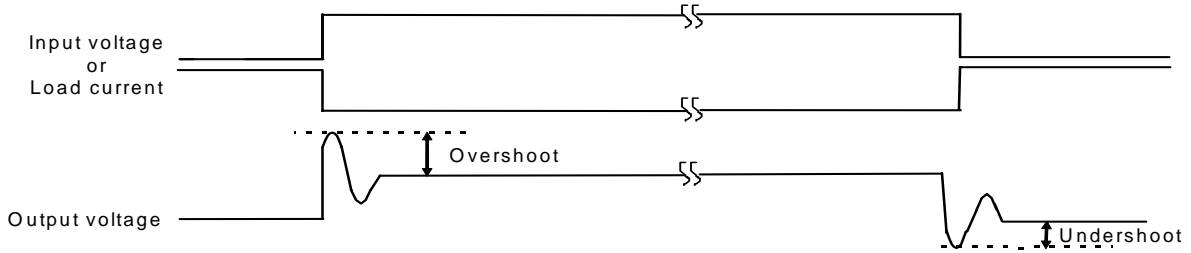
S-818A30A



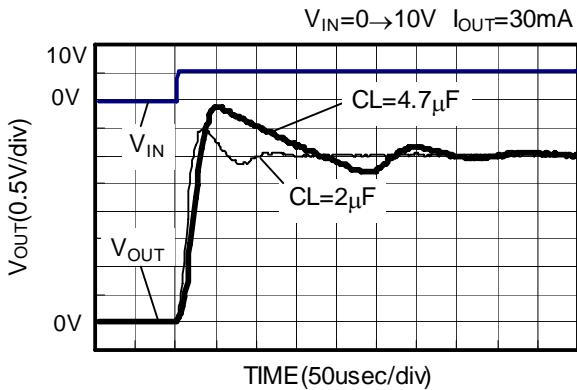
S-818A50A



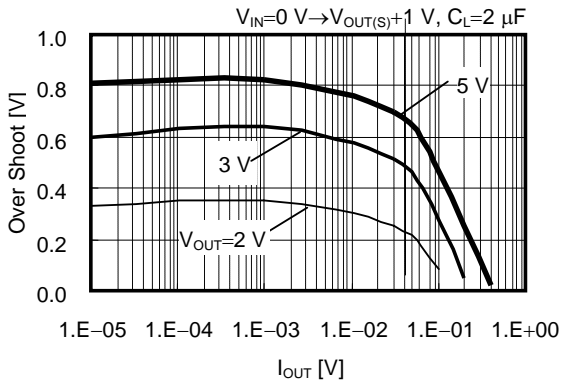
■ **Transient Response Characteristics (S-818A30A, Typical data, Ta=25°C)**



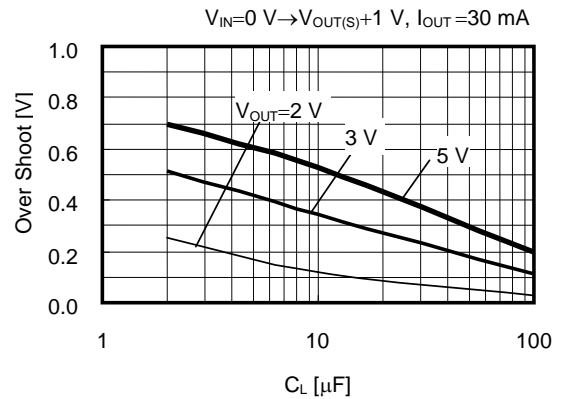
**1. Power on**



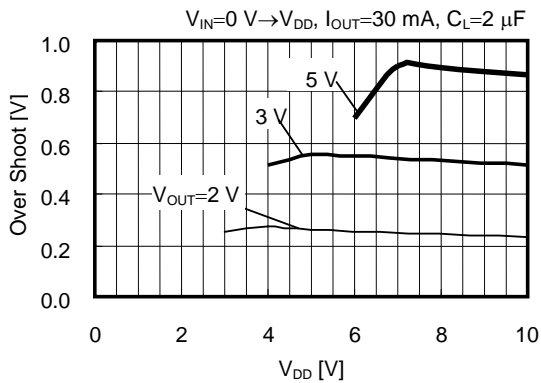
Load dependence of overshoot



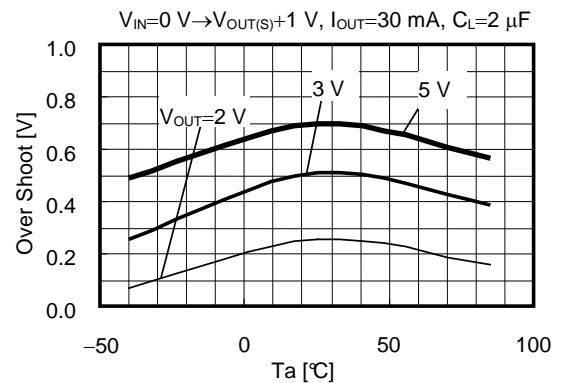
$C_L$  dependence of overshoot



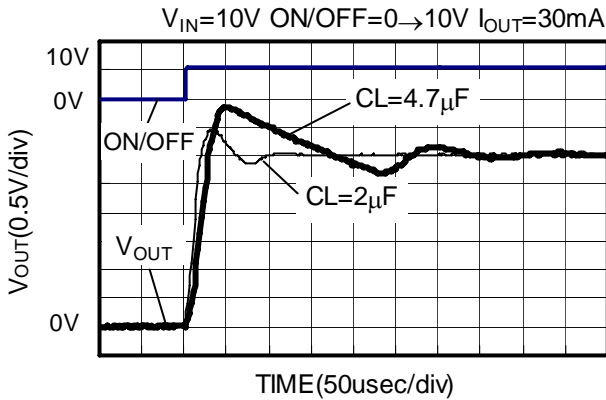
$V_{DD}$  dependence of overshoot



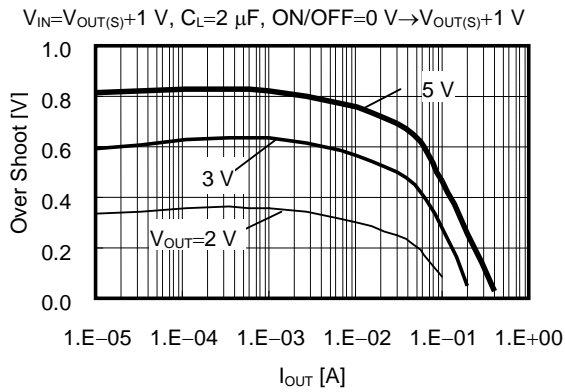
Temperature dependence of overshoot



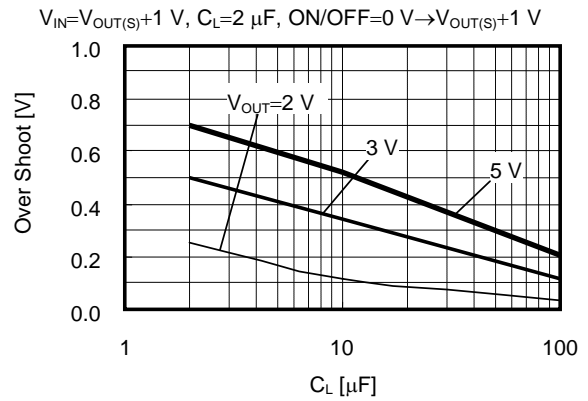
2. Shutdown control



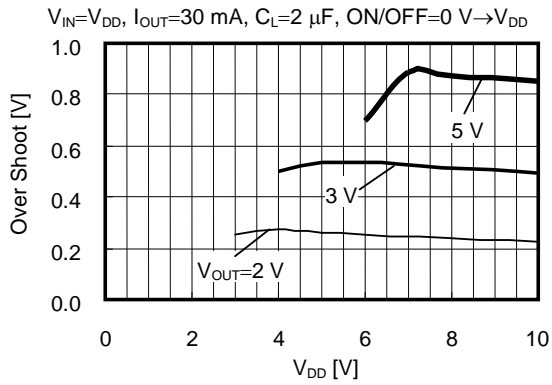
Load dependencies of overshoot



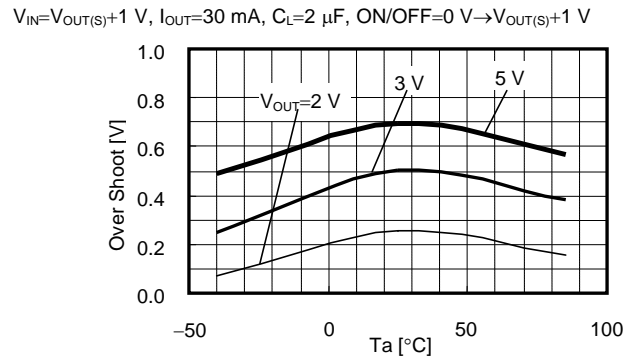
$C_L$  dependence of overshoot



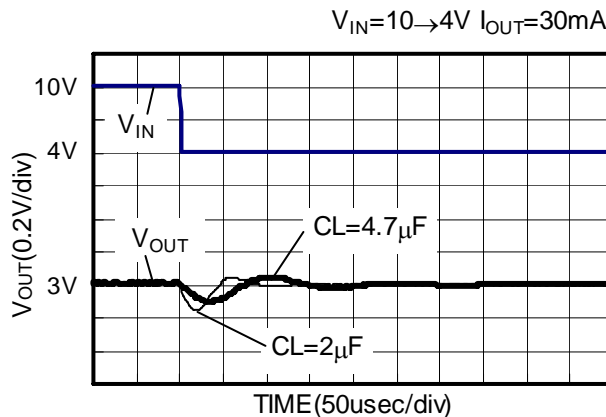
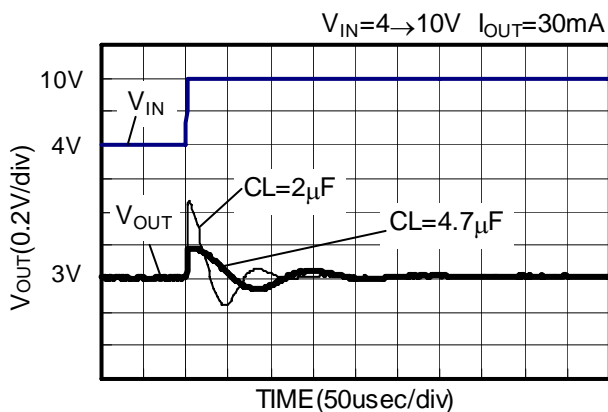
$V_{DD}$  dependencies of overshoot



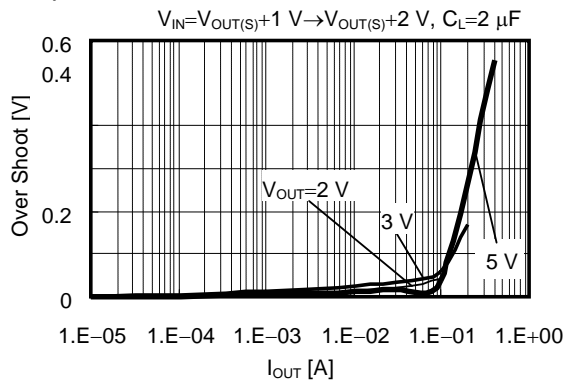
Temperature dependence of overshoot



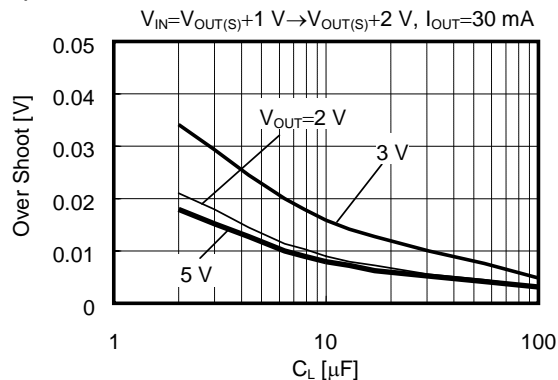
**3. Power fluctuation**



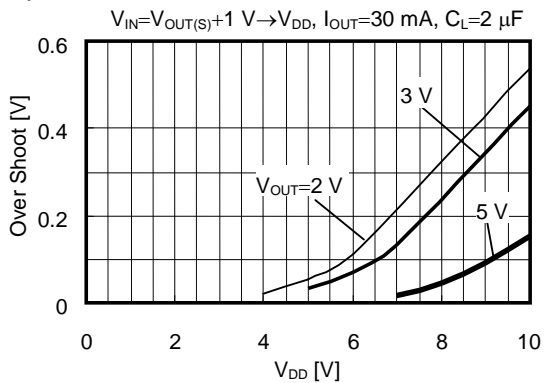
Load dependencies of overshoot



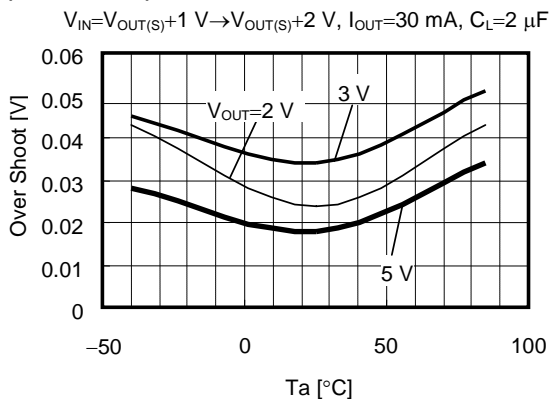
$C_L$  dependence of overshoot



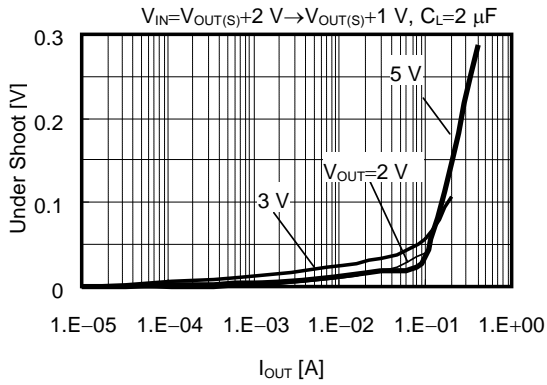
$V_{DD}$  dependencies of overshoot



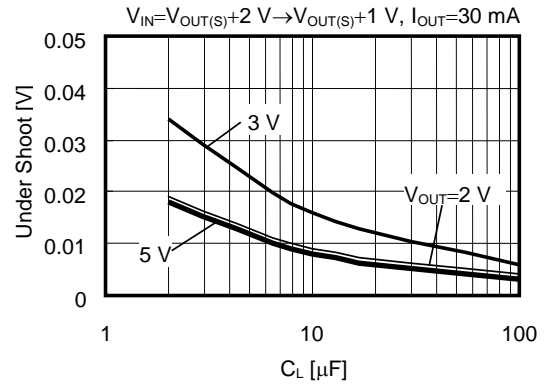
Temperature dependence



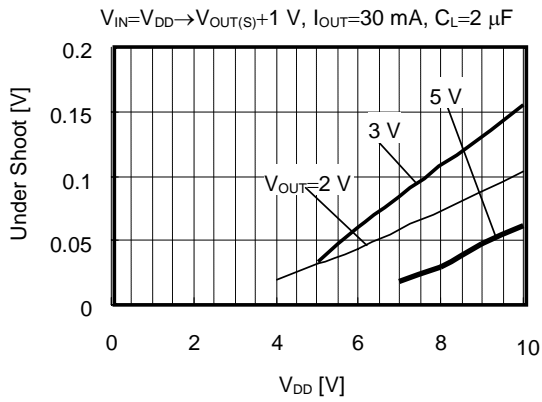
Load dependencies of undershoot



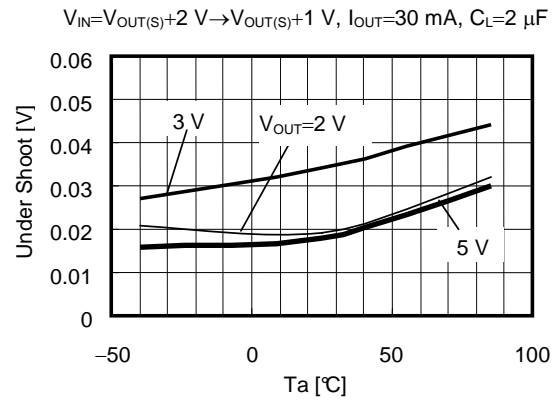
$C_L$  dependence of undershoot



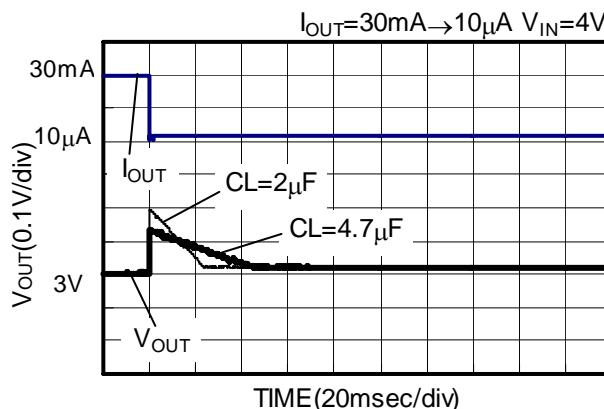
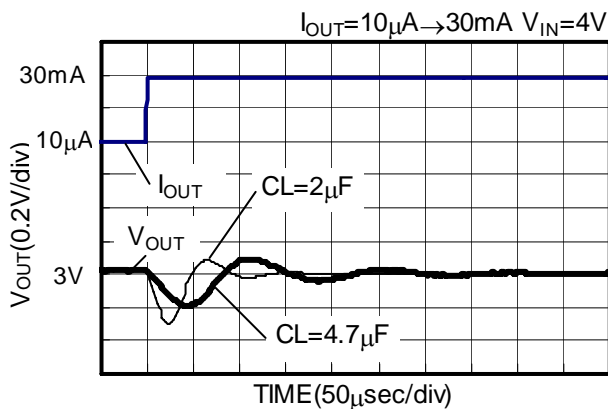
$V_{DD}$  dependencies of undershoot



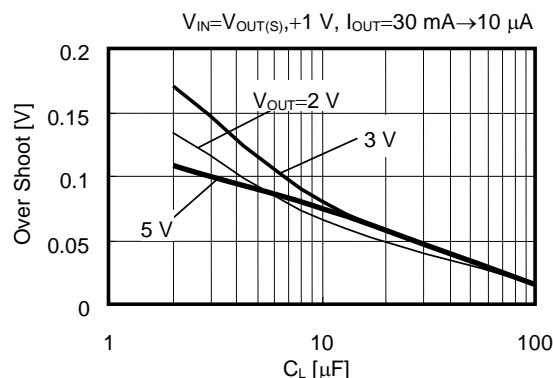
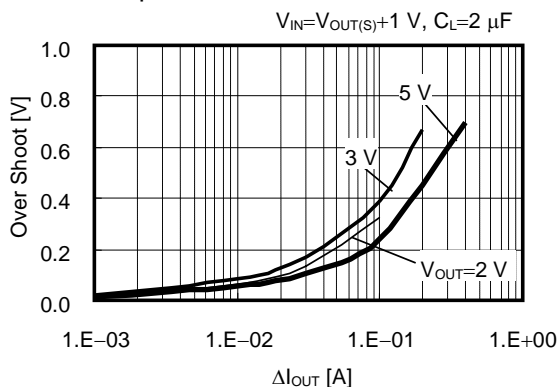
Temperature dependence of undershoot



**4. Load fluctuation**

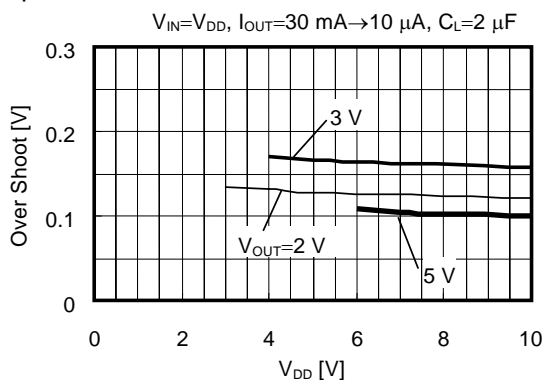


Load current dependence of load fluctuation overshoot  $C_L$  dependence of overshoot

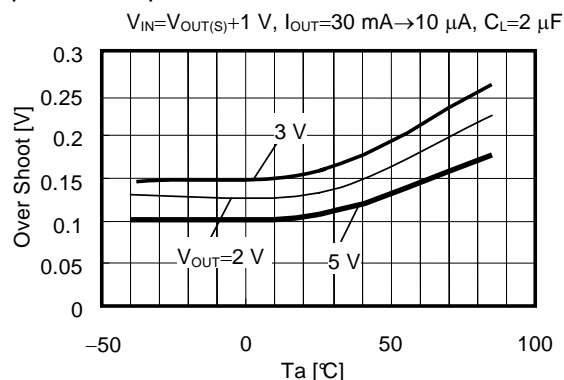


**Remark**  $\Delta I_{OUT}$  shows larger load current at load current fluctuation while smaller current is fixed to 10  $\mu A$ . For example  $\Delta I_{OUT}=1.E-02$  (A) means load current fluctuation from 10 mA to 10  $\mu A$ .

$V_{DD}$  dependencies of overshoot

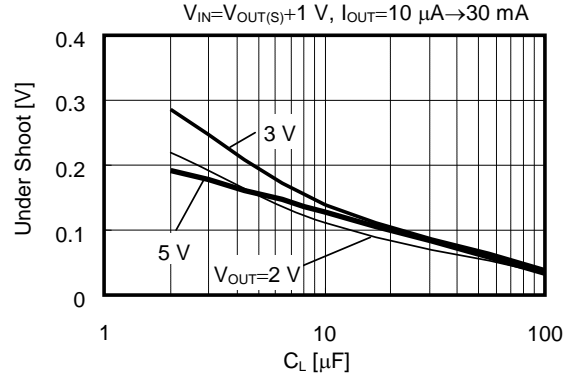
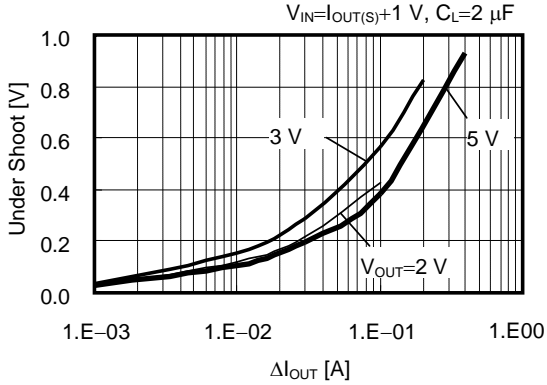


Temperature dependence of overshoot



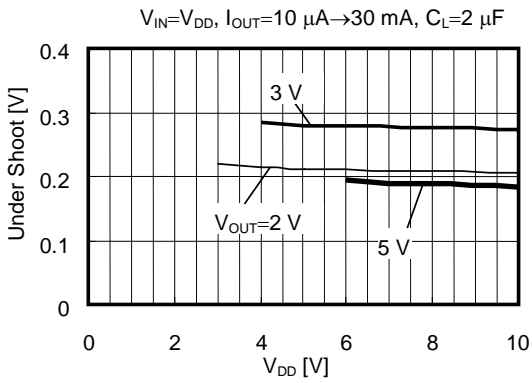


Load current dependence of load fluctuation undershoot  $C_L$  dependence of undershoot

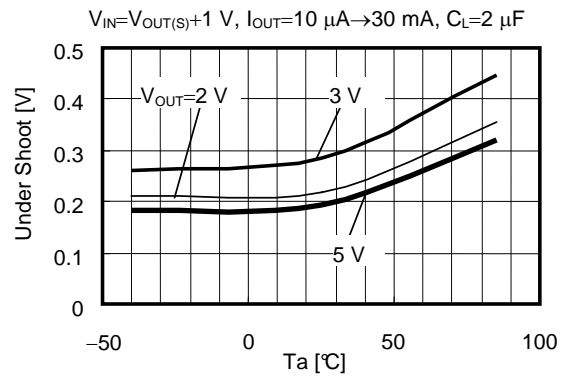


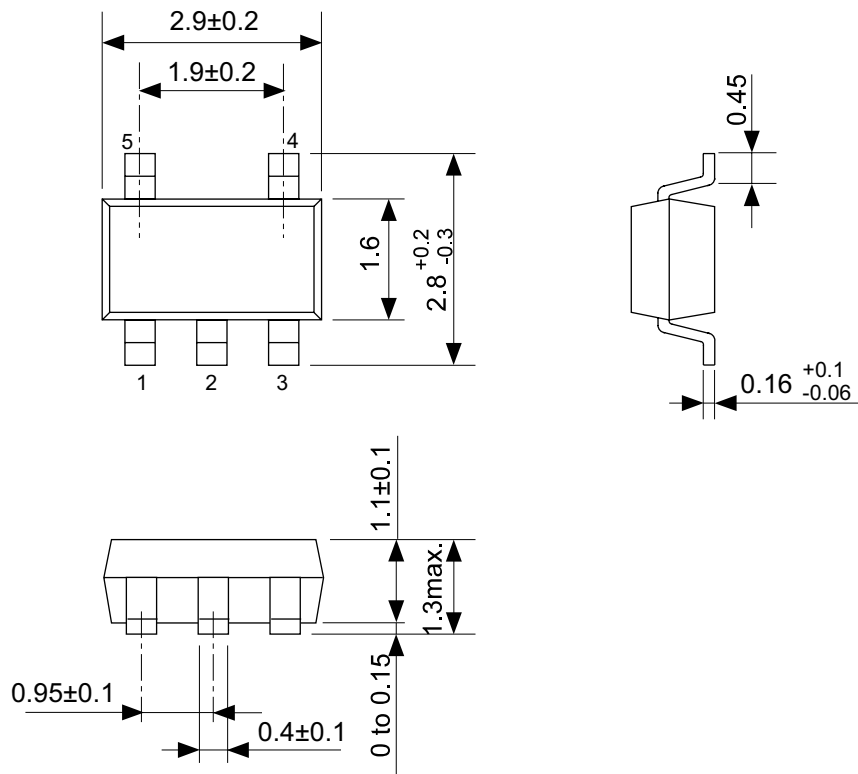
**Remark**  $\Delta I_{OUT}$  shows larger load current at load current fluctuation while smaller current is fixed to 10  $\mu A$ . For example  $\Delta I_{OUT}=1.E-02$  (A) means load current fluctuation from 10  $\mu A$  to 10 mA.

$V_{DD}$  dependence of undershoot



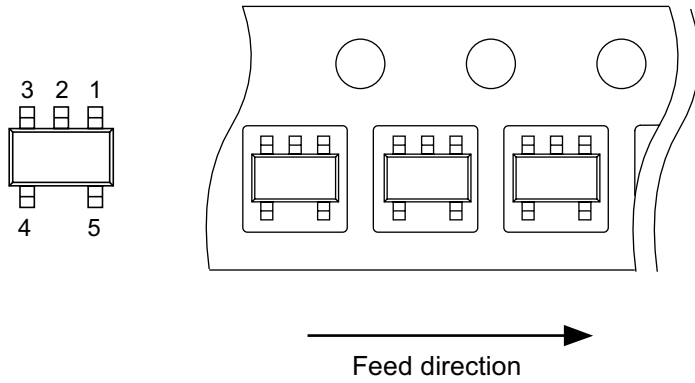
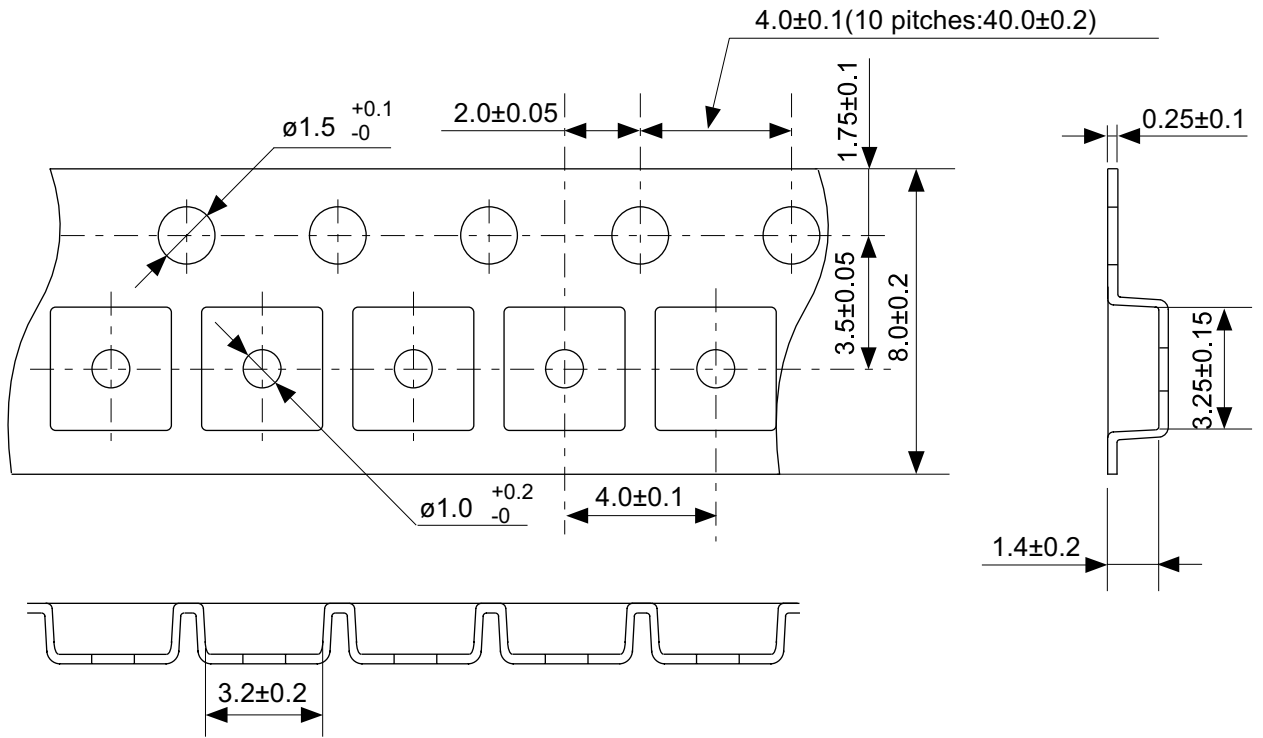
Temperature dependence of undershoot





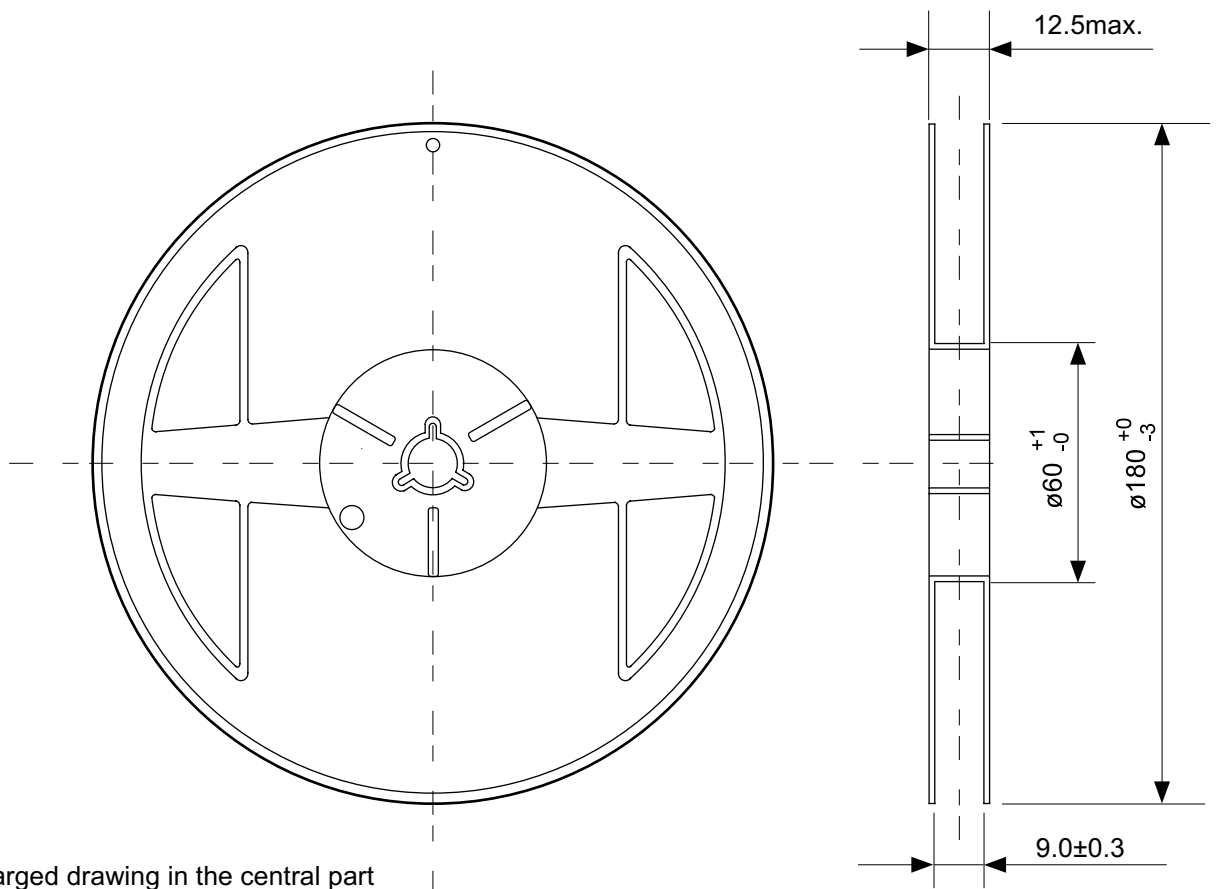
No. MP005-A-P-SD-1.2

TITLE	SOT235-A-PKG Dimensions
No.	MP005-A-P-SD-1.2
SCALE	
UNIT	mm
Seiko Instruments Inc.	

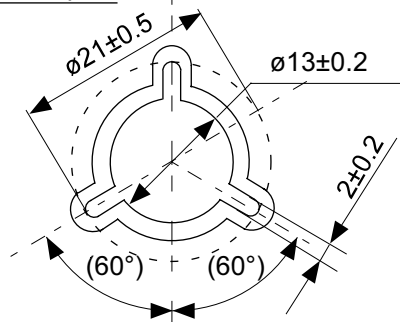


No. MP005-A-C-SD-2.1

TITLE	SOT235-A-Carrier Tape
No.	MP005-A-C-SD-2.1
SCALE	
UNIT	mm
Seiko Instruments Inc.	

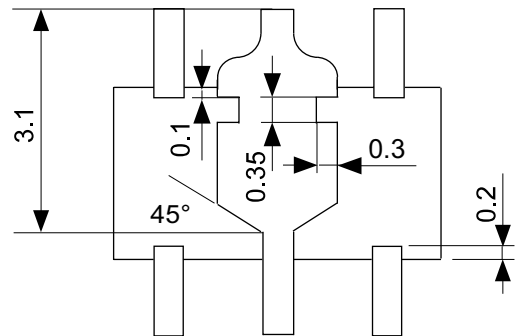
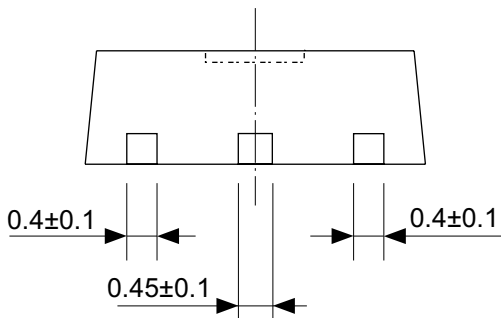
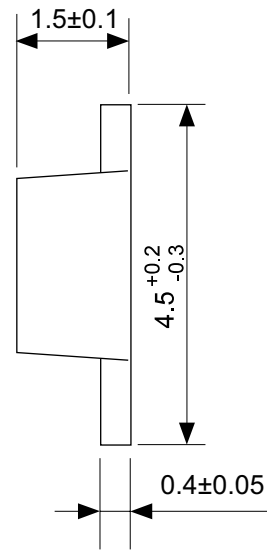
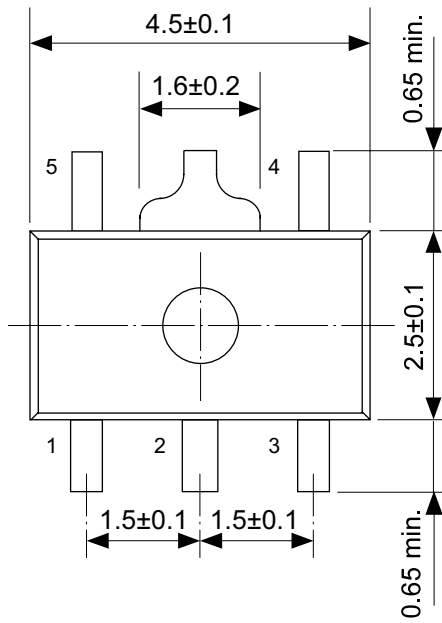


Enlarged drawing in the central part



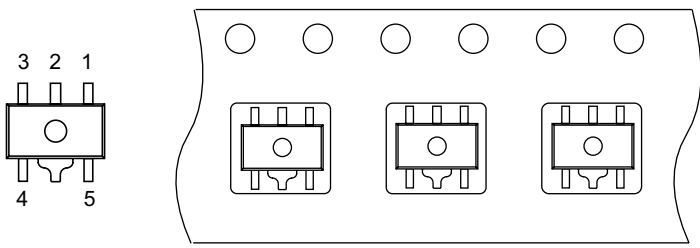
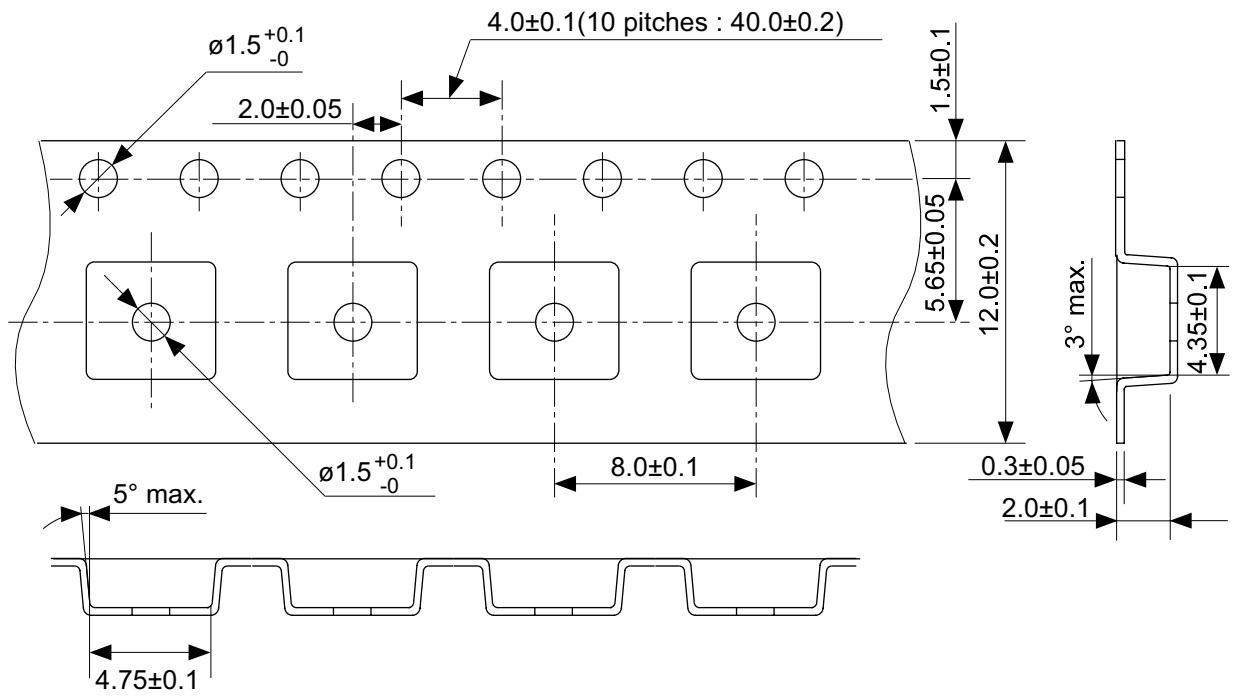
No. MP005-A-R-SD-1.1

TITLE	SOT235-A-Reel		
No.	MP005-A-R-SD-1.1		
SCALE		QTY.	3,000
UNIT	mm		
Seiko Instruments Inc.			



No. UP005-A-P-SD-1.1

TITLE	SOT895-A-PKG Dimensions
No.	UP005-A-P-SD-1.1
SCALE	
UNIT	mm
Seiko Instruments Inc.	

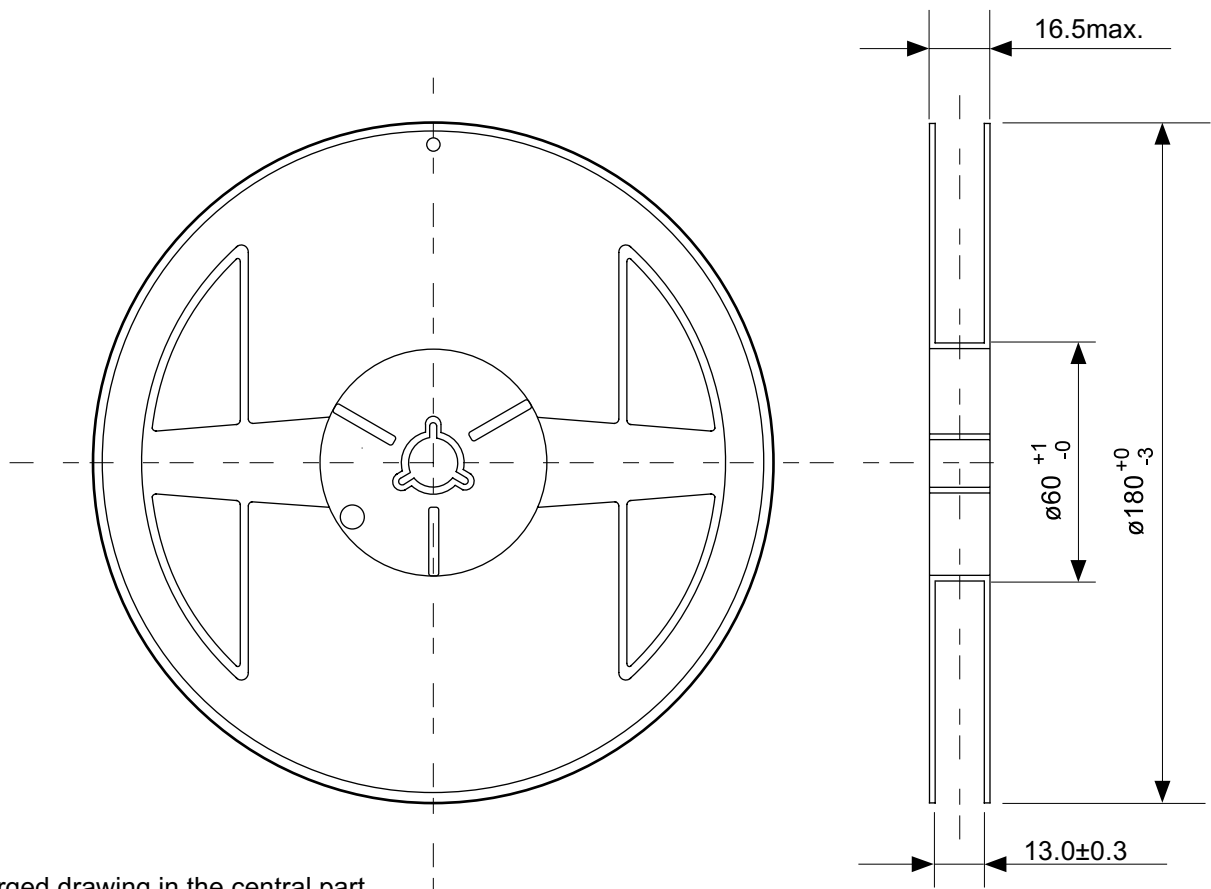


→  
Feed direction

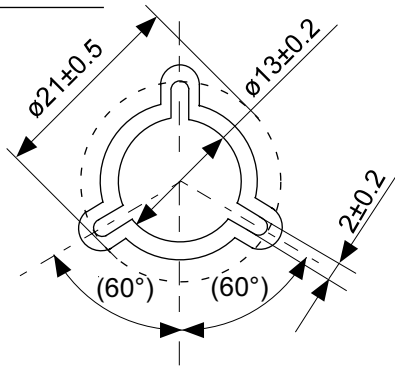
No. UP005-A-C-SD-1.1

TITLE	SOT895-A-Carrier Tape
No.	UP005-A-C-SD-1.1
SCALE	
UNIT	mm

Seiko Instruments Inc.



Enlarged drawing in the central part



No. UP005-A-R-SD-1.1

TITLE	SOT895-A-Reel		
No.	UP005-A-R-SD-1.1		
SCALE		QTY.	1,000
UNIT	mm		
Seiko Instruments Inc.			



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