



### **Dual 150mA LDO Regulator**

### **General Description**

The MIC5210 is a dual linear voltage regulator with very low dropout voltage (typically 10mV at light loads and 140mV at 100mA), very low ground current ( $225\mu$ A at 10mA output), and better than 1% initial accuracy. It also features individual logic-compatible enable/shutdown control inputs.

Both regulator outputs can supply up to 150mA at the same time as long as each regulator's maximum junction temperature is not exceeded.

Key features include an undervoltage monitor with an error flag output, a reference bypass pin to improve its already low-noise performance (8-pin versions only), reversedbattery protection, current limiting, and overtemperature shutdown.

Designed especially for hand-held battery powered devices, the MIC5210 can be switched by a CMOS or TTL compatible logic signal, or the enable pin can be connected to the supply input for 3-terminal operation. When disabled, power consumption drops nearly to zero. Dropout ground current is minimized to prolong battery life.

Key features include current limiting, overtemperature shutdown, and protection against reversed battery.

The MIC5210 is available in 2.7V, 2.8V, 3.0V, 3.3V, 3.6V, 4.0V and 5.0V fixed voltage configurations. Other voltages are available; contact Micrel for details.

Data sheets and support documentation can be found on Micrel's web site at www.micrel.com.

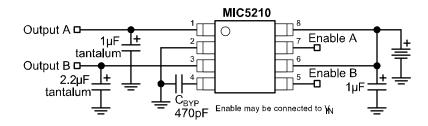
### **Features**

- Micrel Mini 8<sup>™</sup> MSOP package
- Up to 150mA per regulator output
- Low quiescent current
- · Low dropout voltage
- · Wide selection of output voltages
- Tight load and line regulation
- Low temperature coefficient
- Current and thermal limiting
- Reversed input polarity protection
- Zero off-mode current
- Logic-controlled electronic enable

### Applications

- Cellular telephones
- Laptop, notebook, and palmtop computers
- Battery-powered equipment
- Bar code scanners
- SMPS post-regulator/dc-to-dc modules
- High-efficiency linear power supplies

### **Typical Application**



Low-Noise + Ultralow-Noise (Dual) Regulator

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### **Ordering Information**

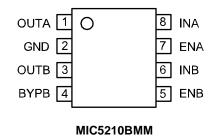
Part Number			Voltage*	A	Junction	Deelkerre		
Standard Marking		Pb-Free Marking		Side A/B	Accuracy	Temp. Range	Package	
MIC5210-2.7BMM		Contact Factory		2.7V/2.7V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-2.8BMM		MIC5210-2.8YMM		2.8V/2.8V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-2.9BMM		Contact Factory		2.9V/2.9V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-3.0BMM		MIC5210-3.0YMM		3.0V/3.0V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-3.3BMM		MIC5210-3.3YMM		3.3V/3.3V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-3.6BMM		Contact Factory		3.6V/3.6V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-4.0BMM		Contact Factory		4.0V/4.0V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-5.0BMM		MIC5210-5.0YMM		5.0V/5.0V	1.0%	–40° to +125°C	8-Pin MSOP	
MIC5210-2.8/3.0BMM	MP	MIC5210-MPYMM**	MPY	2.8V/3.0V	1.0%	–40° to +125°C	8-Pin MSOP	

\* Other voltages available. Contact Micrel for details.

\*\* Order Entry P/N for Pb-Free has been abbreviated in compliance with Micrel systems. MIC5210-MPYMM = Full P/N: MIC5210-2.8/3.0YMM.

Voltage Code	Key
2.8	М
3.0	Р

# **Pin Configuration**



# **Pin Description**

Pin Number	Pin Name	Pin Function	
1	OUTA	Regulator Output A	
2	GND	Ground	
3	OUTB	Regulator Output B	
4	BYPB	eference Bypass B: Connect external 470pF capacitor to GND to reduce utput noise in regulator "B". May be left open.	
5	ENB	Enable/Shutdown B (Input): CMOS compatible input. Logic high = enable, logic low or open = shutdown. Do not leave floating.	
6	INB	Supply Input B	
7	ENA	Enable/Shutdown A (Input): CMOS compatible input. Logic high = enable, logic low or open = shutdown. Do not leave floating.	
8	INA	Supply Input A	

# **Absolute Maximum Ratings**

Supply Input Voltage (V <sub>IN</sub> )	–20V to +20V
Enable Input Voltage (V <sub>EN</sub> )	–20V to +20V
Power Dissipation (P <sub>D</sub> )	Internally Limited
Storage Temperature Range	–60°C to +150°C
Lead Temperature (soldering, 5 sec.)	260°C

# **Operating Ratings**

Supply Input Voltage (V <sub>IN</sub> )	2.5V to 16V
Enable Input Voltage (V <sub>EN</sub> )	0V to 16V
Junction Temperature (T <sub>J</sub> )	–40°C to +85°C
Thermal Resistance $(\theta_{JA})$	Note 1

### **Electrical Characteristics**

1/1 = 1/1 = 11/1 = 1000 = 1000	)μF; V <sub>EN</sub> ≥ 2.0V; T <sub>J</sub> = 25°C, <b>bold</b> values indicate	$\sim 10^{\circ}$ C $\sim$ T $\sim \pm 125^{\circ}$ C uplose poted
$V_{\rm IN} = V_{\rm OUT} + I_V$ . II = IUUUA. CI = I.U	JUF. VEN 2 2.0V. TT - 23 C. <b>DOID</b> VAIUES INVICALE	$= -40 \ \text{C} \times 1 \ \text{F} \times 120 \ \text{C}$ . UHIESS HOLEU.

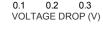
Symbol	Parameter	Condition	Min	Тур	Max	Units
Vo	Output Voltage Accuracy	Variation from specified V <sub>OUT</sub>	-1 -2		1 2	% %
$\Delta V_0 / \Delta T$	Output Voltage Temperature Coefficient	Note 2		40		ppm/°C
$\Delta V_0/V_0$	Line Regulation	$V_{IN} = V_{OUT} + 1V$ to 16V		0.004	0.012 <b>0.05</b>	%/V %/V
$\Delta V_0/V_0$	Load Regulation	I <sub>L</sub> = 0.1mA to 150mA ( <b>Note 3</b> )		0.02	0.2 <b>0.5</b>	% %
$V_{\text{IN}}-V_{\text{O}}$	Dropout Voltage, Note 4	I <sub>L</sub> = 100μA		10	50 <b>70</b>	mV mV
		I <sub>L</sub> = 50mA		110	150 230	mV mV
		I <sub>L</sub> = 100mA		140	250 <b>300</b>	mV mV
		I <sub>L</sub> = 150mA		165	275 <b>350</b>	mV mV
I <sub>GND</sub>	Quiescent Current	$V_{EN} \le 0.4V$ (shutdown) $V_{EN} \le 0.18V$ (shutdown)		0.01	1 5	μΑ μΑ
I <sub>GND</sub>	Ground Pin Current, <b>Note 5</b> (per regulator)	V <sub>EN</sub> ≥ 2.0V, I <sub>L</sub> = 100µA		80	125 <b>150</b>	μA μA
	(P =	I <sub>L</sub> = 50mA		350	600 800	μA μA
		I <sub>L</sub> = 100mA		600	1000 <b>1500</b>	μA μA
		I <sub>L</sub> = 150mA		1300	1900 <b>2500</b>	μΑ μΑ
PSRR	Ripple Rejection	Frequency = 100Hz, I <sub>L</sub> = 100µA		75		dB
I <sub>LIMIT</sub>	Current Limit	V <sub>OUT</sub> = 0V		320	500	mA
$\Delta V_{O} / \Delta P_{D}$	Thermal Regulation	Note 6		0.05		%/W
e <sub>no</sub>	Output Noise (Regulator B only)	$I_L = 50$ mA, $C_L = 2.2\mu$ F, 470pF from BYPB to GND		260		nV/√Hz

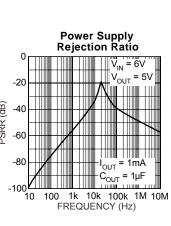
Symbol	Parameter	Condition	Min	Тур	Max	Units
Enable In	put					
V <sub>IL</sub>	Enable Input Logic-Low Voltage	Regulator shutdown			0.4 <b>0.18</b>	V V
V <sub>IH</sub>	Enable Input Logic-High Voltage	Regulator enabled	2.0			V
I <sub>IL</sub> I <sub>IH</sub>	Enable Input Current	$\begin{split} V_{IL} &\leq 0.4V \\ V_{IL} &\leq 0.18V \\ V_{IH} &\geq 2.0V \\ V_{IH} &\geq 2.0V \end{split}$		0.01 5	-1 -2 20 25	μΑ μΑ μΑ μΑ

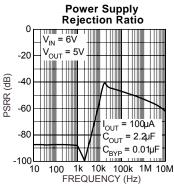
#### Notes:

- 1. Absolute maximum ratings indicate limits beyond which damage to the component may occur. Electrical specifications do not apply when operating the device outside of its operating ratings. The maximum allowable power dissipation is a function of the maximum junction temperature,  $T_{J(max)}$ , the junction-to-ambient thermal resistance,  $\theta_{JA}$ , and the ambient temperature,  $T_A$ . The maximum allowable power dissipation at any ambient temperature is calculated using:  $P_{D(max)} = (T_{J(max)} T_A) / \theta_{JA}$ . Exceeding the maximum allowable power dissipation will result in excessive die temperature, and the regulator will go into thermal shutdown. The  $\theta_{JA}$  of the 8-pin MSOP (MM) is 200°C/W mounted on a PC board (see "Thermal Considerations" section for further details).
- 2. Output voltage temperature coefficient is defined as the worst case voltage change divided by the total temperature range.
- 3. Regulation is measured at constant junction temperature using low duty cycle pulse testing. Parts are tested for load regulation in the load range from 0.1mA to 150mA. Changes in output voltage due to heating effects are covered by the thermal regulation specification.
- 4. Dropout Voltage is defined as the input to output differential at which the output voltage drops 2% below its nominal value measured at 1V differential.
- 5. Ground pin current is the regulator quiescent current plus pass transistor base current. The total current drawn from the supply is the sum of the load current plus the ground pin current.
- Thermal regulation is defined as the change in output voltage at a time "t" after a change in power dissipation is applied, excluding load or line regulation effects. Specifications are for a 150mA load pulse at V<sub>IN</sub> = 16V for t = 10ms.

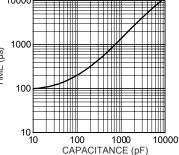
#### **Typical Characteristics** Power Supply Rejection Ratio 0 0 V<sub>IN</sub> = 6V = 5\ ัดมา -20 -20 PSRR (dB) PSRR (dB) -40 -40 -60 -60 -80 -80 100 OUT C<sub>OUT</sub> = 1µF -100 10 1.111000 -100 1k 10k 100k 1M 10M 100 10 FREQUENCY (Hz) Power Supply Rejection Ratio 0 0 = 6V VIN VIN = 51 -20 -20 PSRR (dB) PSRR (dB) -40 -40 -60 -60 100m -80 -80 ουτ C<sub>OUT</sub> = 1µF -100 10 -100 100 1k 10k 100k 1M 10M 10 FREQUENCY (Hz) Power Supply Rejection Ratio 10000 0 /<sub>IN</sub> = 6V = 5\ -20 ้อบา 1000 PSRR (dB) TIME (µs) -40 -60 100 = 10mA OUT C<sub>OUT</sub> = 2.2µF -80 C<sub>BYP</sub> = 0.01µF 10∟ 10 -100 0 1k 10k 100k 1M 10M FREQUENCY (Hz) 10 100 **Power Supply Ripple Rejection** vs. Voltage Drop 100 0 90 IN **RIPPLE REJECTION (dB** 80 -20 1m/ 70 60 PSRR (dB) -40 I<sub>OUT</sub> = 100mA 50 40 -60 10m/ 30 C<sub>OUT</sub> = 2.2µF 20 -80 $C_{BYP} = 0.01 \mu F$ 10 0 0.1 0.2 0.3 0.4 0

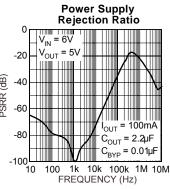


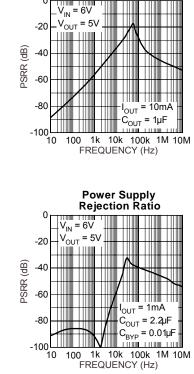




Turn-On Time vs. Bypass Capacitance



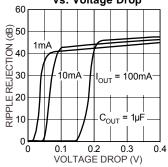


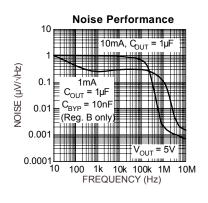


Power Supply Rejection Ratio

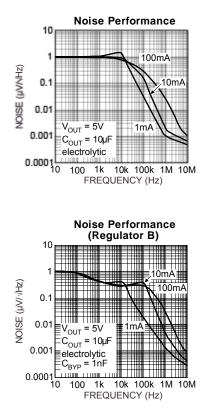
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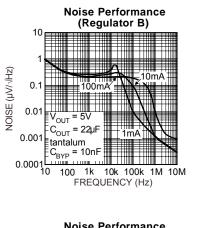
Power Supply Ripple Rejectio vs. Voltage Drop

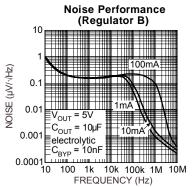


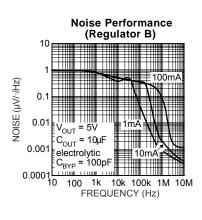


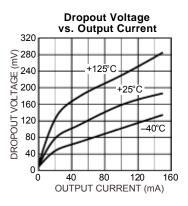
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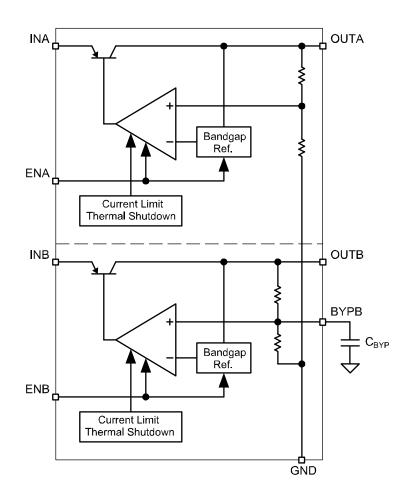








# **Block Diagram**



### **Application Information**

### Enable/Shutdown

Forcing EN (enable/shutdown) high (> 2V) enables the regulator. EN is compatible with CMOS logic gates.

If the enable/shutdown feature is not required, connect EN to IN (supply input).

### Input Capacitor

A  $1\mu$ F capacitor should be placed from IN to GND if there is more than 10 inches of wire between the input and the ac filter capacitor or if a battery is used as the input.

### **Reference Bypass Capacitor**

BYPB (reference bypass) is connected to the internal voltage reference of regulator B. A 470pF capacitor  $(C_{BYP})$  connected from BYPB to GND quiets this reference, providing a significant reduction in output noise.  $C_{BYP}$  reduces the regulator phase margin; when using  $C_{BYP}$ , output capacitors of 2.2µF or greater are generally required to maintain stability.

The start-up speed of the MIC5210 is inversely proportional to the size of the reference bypass capacitor. Applications requiring a slow ramp-up of output voltage should consider larger values of  $C_{\rm BYP}$ . Likewise, if rapid turn-on is necessary, consider omitting  $C_{\rm BYP}$ .

If output noise is not a major concern, omit  $C_{\mbox{\scriptsize BYP}}$  and leave BYPB open.

### **Output Capacitor**

An output capacitor is required between OUT and GND to prevent oscillation. The minimum size of the output capacitor is dependent upon whether a reference bypass capacitor is used.  $1.0\mu$ F minimum is recommended when  $C_{BYP}$  is not used (see Figure 2).  $2.2\mu$ F minimum is recommended when  $C_{BYP}$  is 470pF (see Figure 1). Larger values improve the regulator's transient response. The output capacitor value may be increased without limit.

The output capacitor should have an ESR (effective series resistance) of about  $5\Omega$  or less and a resonant frequency above 1MHz. Ultralow-ESR capacitors may cause a low-amplitude oscillation and/or underdamped transient response. Most tantalum or aluminum electrolytic capacitors are adequate; film types will work, but are more expensive. Since many aluminum electrolytic capacitors have electrolytes that freeze at about  $-30^{\circ}$ C, solid tantalum capacitors are recommended for operation below  $-25^{\circ}$ C.

At lower values of output current, less output capacitance is required for output stability. The capacitor can be reduced to  $0.47\mu$ F for current below 10mA or  $0.33\mu$ F for currents below 1mA.

### **No-Load Stability**

The MIC5210 will remain stable and in regulation with no load (other than the internal voltage divider) unlike many other voltage regulators. This is especially important in CMOS RAM keep-alive applications.

### **Dual-Supply Operation**

When used in dual supply systems where the regulator load is returned to a negative supply, the output voltage must be diode clamped to ground.

### Thermal Considerations

Multilayer boards having a ground plane, wide traces near the pads, and large supply bus lines provide better thermal conductivity.

The MIC5210-xxBMM (8-pin MSOP) has a thermal resistance of 200°C/W when mounted on a FR4 board with minimum trace widths and no ground plane.

PC Board Dielectric	θյΑ	
FR4	200°C/W	

**MSOP** Thermal Characteristics

For additional heat sink characteristics, please refer to Micrel Application Hint 17, "Calculating P.C. Board Heat Sink Area For Surface Mount Packages".

#### Thermal Evaluation Examples

For example, at 50°C ambient temperature, the maximum package power dissipation is:

$$P_{D(max)} = (125^{\circ}C - 50^{\circ}C) / 200^{\circ}C/W$$
  
= 375mW

If the intent is to operate the 5V version from a 6V supply at the full 150mA load for both outputs in a 50°C maximum ambient temperature, make the following calculation:

$$P_{D(each regulator)} = (V_{IN} - V_{OUT}) \times I_{OUT} + (V_{IN} \times I_{GND})$$
$$= (6V - 5V) \times 150mA + (6V \times 2.5mA)$$
$$= 165mW$$

 $P_{D(both regulators)} = 2 regulators \times 165 mW$ 

= 330mW

The actual total power dissipation of 330mW is below the 375mW package maximum, therefore, the regulator can be used.

Note that both regulators cannot always be used at their maximum current rating. For example, in a 5V input to 3.3V output application at 50°C, if one regulator supplies 150mA, the other regulator is limited to a much lower current. The first regulator dissipates:

$$P_D = (5V - 3.3V) 150 + 2.5mA (5V)$$
  
 $P_D = 267.5mW$ 

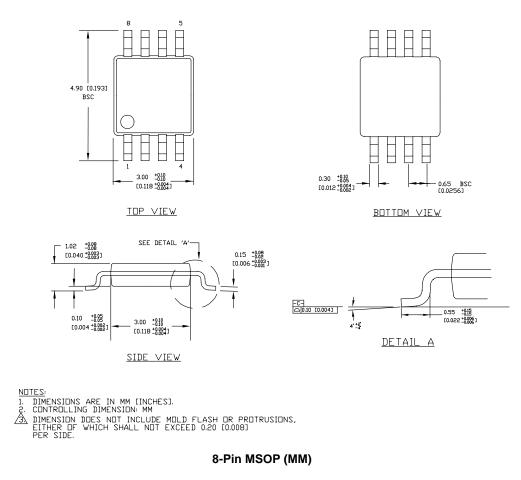
Then, the load that the remaining regulator can dissipate must not exceed:

375mW – 267.5mW = 107.5mW

This means, using the same 5V input and 3.3V output voltage, the second regulator is limited to about 60mA.

Taking advantage of the extremely low-dropout voltage characteristics of the MIC5210, power dissipation can be reduced by using the lowest possible input voltage to minimize the input-to-output voltage drop.

### **Package Information**



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